## HD66717

# (Low-Power Dot-Matrix Liquid Crystal Display Controller/Driver) 

## HITACHI

## Description

The HD66717 dot-matrix liquid crystal display controller and driver LSI displays alphanumerics, katakana, hiragana, and symbols. It can be configured to drive a dot-matrix liquid crystal display under the control of an $\mathrm{I}^{2} \mathrm{C}$ bus, a clock-synchronized serial, or a 4- or 8-bit microprocessor. A single HD66717 is capable of displaying a maximum of four 12-character lines, 40 segments, and 10 annunciators. The HD66717 incorporates all the functions required for driving a dot-matrix liquid crystal display such as display RAM, character generator, and liquid crystal drivers, and a booster for LCD power supply.

The HD66717 provides various functions to reduce the power consumption of an LCD system such as low-voltage operation of 2.4 V or less, a booster for generating a maximum of triple LCD drive voltage from the supplied voltage, and voltage-followers for decreasing the direct current flow in the LCD drive bleeder-resistors. Combining these hardware functions with software functions such as standby and sleep modes allows a fine power control. The HD66717, with the above functions, is suitable for any portable battery-driven product requiring long-term driving capabilities and small size.

## Features

- $5 \times 8$-dot matrix LCD drive
- Four 12-character lines, 40 segments, and 10 annunciators
- Low-power operation support:
- 2.4 to 5.5 V (low voltage)
- Double or triple booster for liquid crystal drive voltage
- Electron volume function and voltage-followers for decreasing the direct current flow in the LCD drive bleeder-resistors
- Standby mode and sleep mode
- Displays up to 10 static annunciators
- $I^{2} \mathrm{C}$ bus or clock-synchronized serial interface; 4- or 8-bit parallel bus interface
- $60 \times 8$-bit display data RAM ( 60 characters max)
- 9,600-bit character generator ROM
- 240 characters ( $5 \times 8$ dots)
- $32 \times 5$-bit character generator RAM
-4 characters ( $5 \times 8$ dots)
- $8 \times 5$-bit segment RAM
- 40 segment-icons and marks max
- 60-segment $\times 34$-common liquid crystal display driver
- Programmable display sizes and duty ratios (see List1)
- Vertical smooth scroll
- Double-height display
- Wide range of instruction functions:
- Display clear, display on/off, icon and mark control, character blink, white-black inverting blinking cursor, icon and mark blink, cursor home, cursor on/off, white-black inverting raster-row
- Hardware reset
- Internal oscillation with an external resistor
- Wide range of LCD drive voltages
-3.0 V to 13.0 V
- Slim chip with/without bump (for COB) and tape carrier package (TCP)


## List 1 Programmable Display Sizes and Duty Ratios

| Display Size | Duty Ratio | Oscillation <br> Frequency | Current <br> Consumption | Multi-plexed-Drive <br> Segments | Static-Drive <br> Annunciators |
| :--- | :--- | :--- | :--- | :--- | :--- |
| 1 line $\times 12$ <br> characters | $1 / 10$ | 40 kHz | $8 \mu \mathrm{~A}$ | 40 | 10 |
| 2 lines $\times 12$ <br> characters | $1 / 18$ | 80 kHz | $15 \mu \mathrm{~A}$ | 40 | 10 |
| 3 lines $\times 12$ <br> characters | $1 / 26$ | 120 kHz | $23 \mu \mathrm{~A}$ | 40 | 10 |
| 4 lines $\times 12$ <br> characters | $1 / 34$ | 160 kHz | $30 \mu \mathrm{~A}$ | 40 | 10 |

Note: Current consumption excludes that for LCD power supply source; $\mathrm{V}_{\mathrm{cc}}=3 \mathrm{~V}$.

## List 2 Ordering Information

| Type Name | External Dimensin | Operation Voltage | Internal Font |
| :--- | :--- | :--- | :--- |
| HD66717A03TA0 | TCP | 2.4 V to 5.5V | Japanese and European fonts |
| HCD66717A03 | Bare chip |  |  |
| HCD66717A03BP | Au-bumped chip |  | Up-side-down pattern of A03 |

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## LCD-II Family Comparison

| Item | $\begin{aligned} & \text { LCD-II } \\ & \text { (HD44780U) } \end{aligned}$ | HD66702R | HD66710 | HD66712U |
| :---: | :---: | :---: | :---: | :---: |
| Power supply voltage | 2.7 V to 5.5V | $\begin{aligned} & \hline 5 \mathrm{~V} \pm 10 \% \text { (standard) } \\ & 2.7 \mathrm{~V} \text { to } 5.5 \mathrm{~V} \\ & \text { (low voltage) } \\ & \hline \end{aligned}$ | 2.7V to 5.5 V | 2.7 V to 5.5 V |
| Liquid crystal drive voltage | 3.0 to 11.0 V | 3.0 V to 8.3V | 3.0 to 13.0 V | 2.7 to 11.0V |
| Maximum display characters per chip | 8 characters $\times$ 2 lines | 20 characters $\times$ 2 lines | 16 characters $\times$ <br> 2 lines/ <br> 8 characters $\times$ <br> 4 lines | 24 characters $\times$ <br> 2 lines/ <br> 12 characters $\times$ <br> 4 lines |
| Segment display | None | None | 40 | 60 (extended to 80) |
| Display duty ratio | $\begin{aligned} & 1 / 8,1 / 11, \text { and } \\ & 1 / 16 \end{aligned}$ | $\begin{aligned} & 1 / 8,1 / 11, \text { and } \\ & 1 / 16 \end{aligned}$ | 1/17 and 1/33 | 1/17 and 1/33 |
| CGROM | 9,920 bits (208 5-x-8 dot characters and $325-x-10$ dot characters) | 7,200 bits (160 5-x-7 dot characters and $325-x-10$ dot characters) | $\begin{aligned} & \hline 9,600 \text { bits } \\ & \text { (240 } 5-\times-8 \text { dot } \\ & \text { characters) } \end{aligned}$ | $\begin{aligned} & \hline 9,600 \text { bits } \\ & \text { (240 } 5-\times-8 \text { dot } \\ & \text { characters) } \end{aligned}$ |
| CGRAM | 64 bytes | 64 bytes | 64 bytes | 64 bytes |
| DDRAM | 80 bytes | 80 bytes | 80 bytes | 80 bytes |
| SEGRAM | None | None | 8 bytes | 16 bytes |
| Segment signals | 40 | 100 | 40 | 60 |
| Common signals | 16 | 16 | 33 | 34 |
| Liquid crystal drive waveform | A | B | B | B |
| Clock source | External resistor or external clock | External resistor or external clock | External resistor or external clock | External resistor or external clock |
| Rf oscillation frequency | $270 \mathrm{kHz} \pm 30 \%$ | $320 \mathrm{kHz} \pm 30 \%$ | $270 \mathrm{kHz} \pm 30 \%$ | $270 \mathrm{kHz} \pm 30 \%$ |
| Liquid crystal voltage booster circuit | None | None | Double or triple booster circuit | Double or triple booster circuit |
| Liquid crystal drive operational amplifier | None | None | None | None |
| Bleeder-resistor for liquid crystal drive | External | External | External | External |
| Liquid crystal contrast adjuster | None | None | None | None |
| Key scan circuit | None | None | None | None |
| Extension driver control signal | Independent control signal | Independent control signal | Used in common with a driver output pin | Independent control signal |
| Reset function | Internal reset circuit | Internal reset circuit | Internal reset circuit | Internal reset circuit or reset input |
| Horizontal smooth scroll | Impossible | Impossible | Dot unit | Dot unit and line unit |
| Vertical smooth scroll | Impossible | Impossible | Impossible | Impossible |
| Number of displayed lines | 1 or 2 | 1 or 2 | 1, 2, or 4 | 1, 2, or 4 |
| Low power control | None | None | Low power mode | Low power mode |
| Bus interface | 4 or 8 bits | 4 or 8 bits | 4 or 8 bits | Serial, 4, or 8 bits |
| Package | 80-pin QFP1420 80-pin TQFP1414 <br> 80-pin bare chip | 144-pin FQFP2020 <br> 144-pin bare chip | $\begin{aligned} & \text { 100-pin QFP1420 } \\ & \text { 100-pin TQFP1414 } \\ & \text { 100-pin bare chip } \\ & \hline \end{aligned}$ | 128-pin TCP <br> 128-pin bare chip |

## LCD-II Family Comparison (cont)

| Item | HD66720 | HD66717 | HD66727 |
| :---: | :---: | :---: | :---: |
| Power supply voltage | 2.7 V to 5.5 V | 2.4 V to 5.5 V | 2.4 V to 5.5 V |
| Liquid crystal drive voltage | 3.0 to 11.0 V | 3.0 to 13.0 V | 3.0 to 13.0 V |
| Maximum display characters per chip | 10 characters $\times$ 1 line/ <br> 8 characters $\times$ <br> 2 lines | 12 characters $\times$ <br> 1 line/2 lines/3 lines/4 lines | 12 characters $\times$ 1 line/2 lines/3 lines/4 lines |
| Segment display | 42 (extended to 80) | 40 (and 10 annunciators) | 40 (and 12 annunciators) |
| Display duty ratio | 1/9 and 1/17 | 1/10, 1/18, 1/26, and 1/34 | 1/10, 1/18, 1/26, and 1/34 |
| CGROM | $\begin{aligned} & 9,600 \text { bits } \\ & (2405-x-8 \text { dot } \\ & \text { characters) } \end{aligned}$ | $\begin{aligned} & \hline 9,600 \text { bits } \\ & (2405-x-8 \text { dot } \\ & \text { characters) } \\ & \hline \end{aligned}$ | $\begin{aligned} & 11,520 \text { bits } \\ & \text { (240 6-x-8 dot } \\ & \text { characters) } \end{aligned}$ |
| CGRAM | 64 bytes | 32 bytes | 32 bytes |
| DDRAM | 40 bytes | 60 bytes | 60 bytes |
| SEGRAM | 16 bytes | 8 bytes | 8 bytes |
| Segment signals | 42 | 60 | 60 |
| Common signals | 17 | 34 | 34 |
| Liquid crystal drive waveform | B | B | B |
| Clock source | External resistor or external clock | External resistor or external clock | External resistor or external clock |
| Rf oscillation frequency | $160 \mathrm{kHz} \pm 30 \%$ | 1 -line mode: $40 \mathrm{kHz} \pm 30 \%$ <br> 2-line mode: $80 \mathrm{kHz} \pm 30 \%$ <br> 3 -line mode: $120 \mathrm{kHz} \pm 30 \%$ <br> 4-line mode: $160 \mathrm{kHz} \pm 30 \%$ | 1-line mode: $40 \mathrm{kHz} \pm 30 \%$ <br> 2-line mode: $80 \mathrm{kHz} \pm 30 \%$ <br> 3-line mode: $120 \mathrm{kHz} \pm 30 \%$ <br> 4-line mode: $160 \mathrm{kHz} \pm 30 \%$ |
| Liquid crystal voltage booster circuit | Double or triple booster circuit | Double or triple booster circuit | Double or triple booster circuit |
| Liquid crystal drive operational amplifier | None | Built-in for each V1 to V5 | Built-in for each V1 to V5 |
| Bleeder-resistor for liquid crystal drive | External | Internal 1/4 and 1/6 bias resistors | Internal 1/4 and 1/6 bias resistors |
| Liquid crystal contrast adjuster | None | Incorporated | Incorporated |
| Key scan circuit | $5 \times 6=30$ keys | None | $4 \times 8=32$ keys |
| Extension driver control signal | Independent control signal | None | None |
| Reset function | Internal reset circuit or reset input | Reset input | Reset input |
| Horizontal smooth scroll | Dot unit and line unit | Impossible | Impossible |
| Vertical smooth scroll | Impossible | Dot (raster-row) unit | Dot (raster-row) unit |
| Number of displayed lines | 1 or 2 | 1, 2, 3, or 4 | 1, 2, 3, or 4 |
| Low power control | Low power mode and sleep mode | Standby mode and sleep mode | Standby mode and sleep mode |
| Bus interface | Serial | $1^{2} \mathrm{C}$, serial, 4 , or 8 bits | $\mathrm{I}^{2} \mathrm{C}$ or clock-synchronized serial |
| Package | 100-pin QFP1420 <br> 100-pin TQFP1414 <br> 100-pin bare chip | Slim chip with/without bumps TCP | Slim chip with/without bumps TCP |

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## HD66717 Block Diagram



## HD66717 Pin Arrangement



## HD66717

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## TCP Dimensions



## HD66717

Pin Functions
Table 1 Pin Functional Description

| Signal | Number of Pins | I/O | Device Interfaced with | Function |
| :---: | :---: | :---: | :---: | :---: |
| IM1, IM0 | 2 | 1 | $\mathrm{V}_{\mathrm{cc}}$ or GND | Selects interface mode with the MPU: <br> IM1, IM0 = GND, GND: $I^{2} C$ bus mode (receive) <br> IM1, IM0 = GND, $\mathrm{V}_{\text {cc }}$ : Clock-synchronized serial mode (receive) <br> IM1, IM0 = $\mathrm{V}_{\mathrm{cc}}$, GND: 8-bit bus mode <br> $\mathrm{IM} 1, \mathrm{IM} 0=\mathrm{V}_{\mathrm{cc}}, \mathrm{V}_{\mathrm{cc}}: 4$-bit bus mode |
| RS/CS* | 1 | 1 | MPU | Selects the HD66717 during clock-synchronized serial mode: <br> Low: HD66717 is selected and can be accessed High: HD66717 is not selected and cannot be accessed Selects the registers during 4 - or 8 -bit bus mode: Low: Instruction register (write); busy flag and address counter (read) <br> High: Data registers (write/read) |
| RW/SDA | 1 | I/O, I | MPU | Inputs serial (receive) data and outputs the acknowledge bit during I ${ }^{2} \mathrm{C}$ bus mode; Inputs serial (receive) data during clock-synchronous serial mode; selects read/write during 4or 8 -bit bus mode: <br> Low: Write <br> High: Read |
| E/SCL | 1 | 1 | MPU | Inputs serial clock pulses during $I^{2} \mathrm{C}$ bus mode and clocksynchronized serial mode; enables data read/write during 4 - or 8 -bit bus mode |
| DB7 DB6, DB5/ID5 DB4/ID4 | 4 | I, I/O | MPU | Inputs the HD66717's identification code (ID5, ID4) during $I^{2} \mathrm{C}$ bus mode and clock-synchronized serial mode;must be fixed to high or low (DB7 and DB6...). <br> Four high-order bidirectional data bus pins for tristate data transfer during 8-bit bus mode. <br> Bidirectional data bus pins during 4-bit bus mode. |
| DB3/ID3, <br> DB2/ID2, <br> DB1/ID1, <br> DB0/IDO, | 4 | I, I/O | MPU | Inputs the HD66717's identification code (ID3 to ID0) during $I^{2} \mathrm{C}$ bus mode and clock-synchronized serial mode; must be fixed to high or low. <br> Four low-order bidirectional data bus pins for tristate data transfer during 8 -bit bus mode. <br> Must be left disconnected during 4-bit bus mode since they are not used. |
| COMS1, COMS2 | 2 | 0 | LCD | Common output signals for segment icon display. |

Table 1 Pin Functional Description (cont)

| Signal | Number of Pins | 1/0 | Device Interfaced with | Function |
| :---: | :---: | :---: | :---: | :---: |
| COM1 to COM32 | 32 | O | LCD | Common output signals for character display: COM1 to COM8 for the first line, COM9 to COM16 for the second line, COM17 to COM24 for the third line, and COM25 to COM32 for the fourth line. All the unused pins output deselection waveforms. During sleep mode (SLP=1) or standby mode ( $\mathrm{STB}=1$ ), all pins output $\mathrm{V}_{\mathrm{cc}}$ level. |
| $\begin{aligned} & \hline \text { SEG1 to } \\ & \text { SEG60 } \end{aligned}$ | 60 | O | LCD | Segment output signals for segment icon display and character display. During sleep mode (SLP = 1) or standby mode (STB = 1), all pins output $\mathrm{V}_{\mathrm{cc}}$ level. |
| ACOM | 1 | O | LCD | Common output signal for annunciator display; can drive display statically between $\mathrm{V}_{\mathrm{cc}}$ and AGND levels; outputs $V_{C C}$ level while annunciator display is turned off ( $D A=0$ ). |
| $\begin{aligned} & \text { ASEG1 to } \\ & \text { ASEG10 } \end{aligned}$ | 10 | O | LCD | Segment output signals for annunciator display; can drive display statically between $\mathrm{V}_{c c}$ and AGND levels; output $\mathrm{V}_{\mathrm{cc}}$ level while annunciator display is turned off ( $\mathrm{DA}=0$ ). |
| V2/V3 | 2 | I | Open or Short-circuited | V2/V3 are voltage levels for the internal operational amplifiers; can drive LCD with $1 / 4$ bias when V2 and V3 are short-circuited and with $1 / 6$ bias when they are left disconnected. |
| V1OUT to V5OUT | 5 | I or O | - | Used for output from the internal operational amplifiers when they are used (OPOFF = GND); when amplifiers' driving capability is insufficient, attach a capacitor to stabilize the output. Especially these capacitors for V1OUT and V4OUT must be attached in 1/26 duty and $1 / 34$ duty. When the amplifiers are not used (OPOFF = $\mathrm{V}_{\mathrm{cc}}$ ); V1 to V 5 voltages can be supplied to these pins externally. |
| VREFP, <br> VREF, <br> VREFM | 3 | I | Open or Short-circuited | Adjusts the driving capability of the internal operational amplifiers according to the LCD power supply voltage. |
|  |  |  |  | $\begin{array}{ll}\text { LCD Power Supply } & \text { Pin Settings VREF, } \\ \text { Voltage }\left(V_{c c}-V_{E E}\right) & \text { VREFP, and VREFM }\end{array}$ |
|  |  |  |  | $\mathrm{V}_{\text {CC }}-\mathrm{V}_{\text {EE }}: 3 \mathrm{~V}-5 \mathrm{~V}$ ( Only VREF and VREFP shorted |
|  |  |  |  | $\mathrm{V}_{\text {CC }}-\mathrm{V}_{\text {EE }}: 4 \mathrm{~V}-6 \mathrm{~V} \quad$ All pins open |
|  |  |  |  | $\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\text {EE }}: 5 \mathrm{~V}-8 \mathrm{~V} \quad$ All pins shorted |
|  |  |  |  | $\mathrm{V}_{\text {CC }}-\mathrm{V}_{\mathrm{EE}}$ : 7 V or more Only VREF and VREFM shorted |
| $\mathrm{V}_{\mathrm{EE}}$ | 2 | - | Power supply | GND power supply for LCD drive $\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}=13 \mathrm{~V}$ max. |
| $\mathrm{V}_{\text {cc }} / \mathrm{GND}$ | 10 | - | Power supply | $\mathrm{V}_{\text {cc }}$ : +2.4 V to +5.5V, GND (logic): 0 V |
| AGND | 2 | - | Power supply | Low level power supply for annunciator display; can adjust contrast of annunciators; AGND ${ }^{3}$ GND. |
| $\begin{aligned} & \hline \text { OSC1/ } \\ & \text { OSC2 } \end{aligned}$ | 2 | - | Oscillation resistor/clock | For R-C oscillation, connect an external resistor For external clock supply, input clock pulses to OSC1. |

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Table 1 Pin Functional Description (cont)

| Signal | Number of Pins | I/O | Device Interfaced with | Function |
| :---: | :---: | :---: | :---: | :---: |
| Vci | 2 | I | Power supply | Inputs a reference voltage and supplies power to the booster; generates the liquid crystal display drive voltage from the operating voltage. |
| V50UT2 | 1 | O | $V_{\text {EE }}$ pin/ Booster capacitance | Voltage input to the Vci pin is boosted twice and output. When the voltage is boosted three times, the same capacitance as that of C1-C2 should be connected here. |
| V50UT3 | 1 | O | $\mathrm{V}_{\text {EE }} \mathrm{pin}$ | Voltage input to the Vci pin is boosted three times and output. |
| C1/C2 | 2 | - | Booster capacitance | External capacitance should be connected here when using the booster. |
| RESET* | 1 | I | - | Reset pin. Initializes the LSI when low. |
| EXM | 1 | I | MPU | External alternating signal used for annunciator display during standby mode. If annunciator display is not used, EXM must be fixed to $V_{c c}$ or GND. |
| $\overline{\text { SFT }}$ | 1 | 1 | $\mathrm{V}_{\mathrm{cc}}$ or GND | Selects the SEG output pin arrangement: when SFT = GND, SEG1 is connected to the far left of the LCD panel and when SFT $=\mathrm{V}_{\mathrm{c}}$, SEG60 is connected to the far left of the LCD panel |
| OPOFF | 1 | I | $\mathrm{V}_{\mathrm{cc}}$ or GND | Turns the internal operational amplifier off when OPOFF = $\mathrm{V}_{\mathrm{cc}}$, and turns it on when OPOFF = GND. If the amplifier is turned off (OPOFF $=\mathrm{V}_{\mathrm{cc}}$ ), V 1 to V 5 must be supplied to the V1OUT to V50UT pins. |
| TEST | 1 | I | GND | Test pin. Must be grounded. |

## Block Function Description

## System Interface

The HD66717 has four types of system interfaces: $I^{2} \mathrm{C}$ bus, clock-synchronized serial, 4-bit bus, and 8-bit bus. The interface mode is selected by the IM1 and IM0 pins.

The HD66717 has two 8-bit registers: an instruction register (IR) and a data register (DR).
The IR stores instruction codes, such as display clear, return home, and display control, and address information for the display data RAM (DDRAM), the character generator RAM (CGRAM), and the segment RAM (SEGRAM). The IR can only be written to by MPU and cannot be read from.

The DR temporarily stores data to be written into DDRAM, CGRAM, SEGRAM, or annunciator. Data written into the DR from the MPU is automatically written into DDRAM, CGRAM, SEGRAM, or annunciator by an internal operation. The DR is also used for data storage when reading data from DDRAM, CGRAM, or SEGRAM. When address information is written into the IR, data is read and then stored into the DR from DDRAM, CGRAM, or SEGRAM by an internal operation. Data transfer between the MPU is then completed when the MPU reads the DR. After the read, data in DDRAM, CGRAM, or SEGRAM at the next address is sent to the DR for the next read from the MPU.

These two registers can be selected by the register select (RS) signal in the $4 / 8$-bit bus interface, and by the RS bit in $\mathrm{I}^{2} \mathrm{C}$ bus or clock-synchronized serial interface (Table 2).

## Busy Flag (BF)

When the busy flag is 1 , the HD66717 is in the internal operation mode, and the next instruction will not be accepted. When $\mathrm{RS}=$ low and $\mathrm{R} / \mathrm{W}=$ high in $4 / 8$-bit bus mode (Table 2), the busy flag is output from DB7. The next instruction must be written after ensuring that the busy flag is 0 . The busy flag cannot be read in $I^{2} \mathrm{C}$ bus mode or clock-synchronized serial mode; data must be transferred in appropriate timing considering instruction execution times.

## Address Counter (AC)

The address counter (AC) assigns addresses to DDRAM, CGRAM, or SEGRAM. When the address set instruction is written into the IR, the address information is sent from the IR to the AC. Selection of DDRAM, CGRAM, and SEGRAM is also determined concurrently by the instruction.

After writing into (reading from) DDRAM, CGRAM, or SEGRAM, the AC is automatically incremented by 1 (or decremented by 1). The AC contents are then output to DB0 to DB6 when $\mathrm{RS}=$ low and $\mathrm{R} / \mathrm{W}=$ high in 4/8-bit bus mode (Table 2).

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Table 2 Register Selection

| RS | R/W | Operation |
| :--- | :--- | :--- |
| 0 | 0 | IR write as an internal operation (display clear, etc.) |
| 0 | 1 | Read busy flag (DB7) read and address counter (DB0 to DB6) (4/8-bit bus interface) |
| 1 | 0 | DR write as an internal operation (DR to DDRAM, CGRAM, SEGRAM, or annunciator) |
| 1 | 1 | DR read as an internal operation (DDRAM, CGRAM, or SEGRAM to DR) <br> (4/8-bit bus interface) |

## Display Data RAM (DDRAM)

Display data RAM (DDRAM) stores display data represented in 8 -bit character codes. Its capacity is $60 \times$ 8 bits, or 60 characters, which is equivalent to an area of 12 characters $\times 5$ lines. Any number of display lines (LCD drive duty ratio) from 1 to 4 can be selected by software. Here, assignment of DDRAM addresses is the same for all display modes (Table 3). The line to be displayed at the top of the display (display-start line) can also be selected by register settings. See Table 4.


Figure 1 Address Counter and DDRAM Address
Table 3 DDRAM Addresses and Display Positions

| Display Line | 1st Char. | 2nd <br> Char. | 3rd Char. | 4th <br> Char. | 5th <br> Char. | 6th Char. | 7th Char. | 8th Char. | 9th Char. | 10th <br> Char. | 11th <br> Char. | 12th <br> Char. |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1st | 00 | 01 | 02 | 03 | 04 | 05 | 06 | 07 | 08 | 09 | 0A | OB |
| 2nd | 10 | 11 | 12 | 13 | 14 | 15 | 16 | 17 | 18 | 19 | 1A | 1B |
| 3rd | 20 | 21 | 22 | 23 | 24 | 25 | 26 | 27 | 28 | 29 | 2A | 2B |
| 4th | 30 | 31 | 32 | 33 | 34 | 35 | 36 | 37 | 38 | 39 | 3A | 3B |
| 5th | 40 | 41 | 42 | 43 | 44 | 45 | 46 | 47 | 48 | 49 | 4A | 4B |

Note: Char. indicates character position.

Table 4 Display-Line Modes, Display-Start Line, and DDRAM Addresses
Display-Start Lines

| Display- <br> Line Mode | Duty Ratio | Common Pins | 1st Line (SN = 000) | 2nd Line $(\mathrm{SN}=001)$ | 3rd Line (SN = 010) | 4th Line (SN = 011) | 5th Line (SN = 100) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \hline \text { 1-line } \\ & (\mathrm{NL}=00) \end{aligned}$ | 1/10 | COM1-COM8 | 00H-0BH | 10H-1BH | 20H-2BH | 30H-3BH | 40H-4BH |
| $\begin{aligned} & \hline \text { 2-line } \\ & (\mathrm{NL}=01) \end{aligned}$ | 1/18 | $\begin{aligned} & \text { COM1-COM8 } \\ & \text { COM9-COM16 } \end{aligned}$ | $\begin{aligned} & \hline 00 \mathrm{H}-0 \mathrm{BH} \\ & 10 \mathrm{H}-1 \mathrm{BH} \end{aligned}$ | $\begin{aligned} & 10 \mathrm{H}-1 \mathrm{BH} \\ & 20 \mathrm{H}-2 \mathrm{BH} \end{aligned}$ | $\begin{aligned} & 20 \mathrm{H}-2 \mathrm{BH} \\ & 30 \mathrm{H}-3 \mathrm{BH} \end{aligned}$ | $\begin{aligned} & 30 \mathrm{H}-3 \mathrm{BH} \\ & 40 \mathrm{H}-4 \mathrm{BH} \end{aligned}$ | $\begin{aligned} & 40 \mathrm{H}-4 \mathrm{BH} \\ & 00 \mathrm{H}-0 \mathrm{BH} \end{aligned}$ |
| $\begin{aligned} & \text { 3-line } \\ & (\mathrm{NL}=10) \end{aligned}$ | 1/26 | COM1-COM8 COM9-COM16 COM17-COM24 | $\begin{aligned} & 00 \mathrm{H}-0 \mathrm{BH} \\ & 10 \mathrm{H}-1 \mathrm{BH} \\ & 20 \mathrm{H}-2 \mathrm{BH} \end{aligned}$ | $\begin{aligned} & 10 \mathrm{H}-1 \mathrm{BH} \\ & 20 \mathrm{H}-2 \mathrm{BH} \\ & 30 \mathrm{H}-3 \mathrm{BH} \end{aligned}$ | $\begin{aligned} & 20 \mathrm{H}-2 \mathrm{BH} \\ & 30 \mathrm{H}-3 \mathrm{BH} \\ & 40 \mathrm{H}-4 \mathrm{BH} \end{aligned}$ | $\begin{aligned} & 30 \mathrm{H}-3 \mathrm{BH} \\ & 40 \mathrm{H}-4 \mathrm{BH} \\ & 00 \mathrm{H}-0 \mathrm{BH} \end{aligned}$ | $\begin{aligned} & 40 \mathrm{H}-4 \mathrm{BH} \\ & 00 \mathrm{H}-0 \mathrm{BH} \\ & 10 \mathrm{H}-1 \mathrm{BH} \end{aligned}$ |
| $\begin{aligned} & \hline 4 \text {-line } \\ & (N L=11) \end{aligned}$ | 1/34 | COM1-COM8 COM9-COM16 COM17-COM24 COM25-COM32 | $\begin{aligned} & 00 \mathrm{H}-0 \mathrm{BH} \\ & 10 \mathrm{H}-1 \mathrm{BH} \\ & 20 \mathrm{H}-2 \mathrm{BH} \\ & 30 \mathrm{H}-3 \mathrm{BH} \end{aligned}$ | $\begin{aligned} & 10 \mathrm{H}-1 \mathrm{BH} \\ & 20 \mathrm{H}-2 \mathrm{BH} \\ & 30 \mathrm{H}-3 \mathrm{BH} \\ & 40 \mathrm{H}-4 \mathrm{BH} \end{aligned}$ | $\begin{aligned} & 20 \mathrm{H}-2 \mathrm{BH} \\ & 30 \mathrm{H}-3 \mathrm{BH} \\ & 40 \mathrm{H}-4 \mathrm{BH} \\ & 00 \mathrm{H}-0 \mathrm{BH} \end{aligned}$ | $\begin{aligned} & 30 \mathrm{H}-3 \mathrm{BH} \\ & 40 \mathrm{H}-4 \mathrm{BH} \\ & 00 \mathrm{H}-0 \mathrm{BH} \\ & 10 \mathrm{H}-1 \mathrm{BH} \end{aligned}$ | $\begin{aligned} & 40 \mathrm{H}-4 \mathrm{BH} \\ & 00 \mathrm{H}-0 \mathrm{BH} \\ & 10 \mathrm{H}-1 \mathrm{BH} \\ & 20 \mathrm{H}-2 \mathrm{BH} \end{aligned}$ |

## Character Generator ROM (CGROM)

The character generator ROM generates $5 \times 8$-dot character patterns from 8 -bit character codes (Table 5). It can generate $2405 \times 8$-dot character patterns. User-defined character patterns are also available using a mask-programmed ROM (see the Modifying Character Patterns section.)

## Character Generator RAM (CGRAM)

The character generator RAM of $32 \times 5$ bits allows the user to redefine the character patterns for user fonts. In the case of $5 \times 8$-dot characters, up to four fonts may be redefined.

Write the character codes at addresses 00 H to 03 H into DDRAM to display the character patterns stored in CGRAM.

## Segment RAM (SEGRAM)

The segment RAM is used to enable control of segments such as an icon and a mark by the user program. Segments and characters are driven by a multiplexing drive method.

SEGRAM has a capacity of $8 \times 5$ bits, for controlling the display of a maximum of 40 icons and marks. While COMS1 and COMS2 outputs are being selected, SEGRAM is read and segments (icons and marks) are displayed by a multiplexing drive method ( 20 segments each during COMS1 and COMS2 selection).

Bits in SEGRAM corresponding to segments to be displayed are directly set by the MPU, regardless of the contents of DDRAM and CGRAM.

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## Timing Generation Circuit

The timing generation circuit generates timing signals for the operation of internal circuits such as DDRAM, CGROM, CGRAM, and SEGRAM. RAM read timing for display and internal operation timing by MPU access are generated separately to avoid interfering with each other. Therefore, when writing data to DDRAM, for example, there will be no undesirable interferences, such as flickering, in areas other than the display area.

## Cursor/Blink Control Circuit

The cursor/blink (or white-black inversion) control is used to produce a cursor or a flashing area on the display at a position corresponding to the location stored in the address counter (AC).

For example (Figure 2), when the address counter is 08 H , a cursor is displayed at a position corresponding to DDRAM address (08)H.

## Multiplexing Liquid Crystal Display Driver Circuit

The multiplexing liquid crystal display driver circuit consists of 34 common signal drivers (COM1 to COM32, COMS1, COMS2) and 60 segment signal drivers (SEG1 to SEG60). When the number of lines are selected by a program, the required common signal drivers automatically output drive waveforms, while the other common signal drivers continue to output deselection waveforms.

Character pattern data is sent serially through a 60-bit shift register and latched when all needed data has arrived. The latched data then enables the segment signal drivers to generate drive waveform outputs.

The shift direction of 60-bit data can be selected by the SFT pin; select the direction appropriate to the device mounting configuration.

When multiplexing drive is not used, or during standby or sleep mode, all common and segment signal drivers output the $\mathrm{V}_{\mathrm{CC}}$ level, halting display.

## Annunciator Driver Circuit

The static annunciator drivers, which are specially used for displaying icons and marks, consists of 1 common signal driver (ACOM) and 10 segment signal drivers (ASEG1 to ASEG10). Since this driver circuit operates at the logic operating voltage ( $\mathrm{V}_{\mathrm{cc}}-\mathrm{AGND}$ ), the LCD drive power supply circuit is not necessary, and low-power consumption can be achieved. It is suitable for mark indication during system standby because of its drive capability during standby and sleep modes. When multiplexing drive is not used, or during standby or sleep mode, all common and segment signal drivers output the $\mathrm{V}_{\mathrm{CC}}$ level, halting display.

## Booster

The booster doubles or triples a voltage input to the Vci pin. With this function, both the internal logic units and LCD drivers can be controlled with a single power supply.

## Oscillator

The HD66717 can provide R-C oscillation simply by adding an external oscillation resistor between the OSC1 and OSC2 pins. The appropriate oscillation frequency for operating voltage, display size, and frame frequency can be obtained by adjusting the external-resistor value. Clock pulses can also be supplied externally. Since R-C oscillation is halted during standby mode, current consumption can be reduced.

## V-Pin Voltage-Followers

A voltage-follower for each voltage level (V1 to V5) reduces current consumption by the LCD drive power supply circuit. No external resistors are required because of the internal bleeder-resistor, which generates different levels of LCD drive voltage. The voltage-followers can be turned off while multiplexing drive is not being used.

## Contrast-Adjuster

The contrast-adjuster can adjust LCD contrast by varying LCD drive voltage by software. This function is suitable for selecting appropriate brightness of the LCD or for temperature compensation.


Note: The cursor/blink or white-black inversion control is also active when the address counter indicates the CGRAM or SEGRAM. However, it has no effect on the display.

Figure 2 Cursor Position and DDRAM Address

Table 5 Relation between Character Codes and Character Patterns (ROM code: A03)

|  | 月 UnF FGE - Ene |
| :---: | :---: |
|  |  |
|  |  |
|  | 1HSCSEEMBATTFE |
|  |  |
|  |  |
|  |  |
|  |  |
|  | E8H2ncerm |
|  | 959145EL- |
|  | c* |
|  |  |
|  |  |
|  |  |
|  |  |
|  |  |

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Table 6 Relation between Character Codes and Character Patterns (ROM code: A13)

| $\begin{array}{ll} \substack{\text { Lower } \\ \text { bits }} & \begin{array}{c} \text { Upper } \\ \text { bits } \end{array} \\ \hline \text { and } \end{array}$ | 0000 | 0001 | 0010 | 0011 | 0100 | 0101 | 0110 | 0111 | 1000 | 1001 | 1010 | 1011 | 1100 | 1101 | 1110 | 1111 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| xxxx 0000 | $\left.\begin{gathered} \text { CG } \\ \text { RAM } \\ 1(1) \end{gathered} \right\rvert\,$ |  |  |  |  | \%ione |  |  |  | $0$ |  | -00* | $\because$ | $\cdots \circ c^{\circ}$ |  | ${ }^{\text {b }}$ |
| xxxx 0001 | $\left.\begin{gathered} \text { CG } \\ \text { RAM } \\ (2) \end{gathered} \right\rvert\,$ |  |  |  | Eooe | $0_{0}^{\circ \circ} 0^{\circ \circ} 0^{\circ}$ |  | $\bullet_{6}^{6}$ |  |  | 88 | $\stackrel{\circ}{8}$ |  |  | ${ }^{\circ}$ | ${ }^{\circ} 0{ }^{\circ} \mathrm{E}$ |
| xxxx 0010 | $\begin{gathered} \mathrm{CG} \\ \text { RAM } \\ (3) \\ \hdashline-2 \end{gathered}$ |  | \%: | $\left\|\begin{array}{c\|c} \because \because \bullet 0 \\ \bullet \bullet \\ \hdashline & 8 \end{array}\right\|$ | $0^{\circ 000}$ | $\left.\right\|_{8} ^{8}$ |  |  | $\mid$ |  | \% |  |  |  | - | ${ }^{000}$ |
| xxxx 0011 | $\begin{array}{\|c\|} \hline \text { CG } \\ \text { RAM } \\ \hline(4) \\ \hline-2 \end{array}$ | $\begin{aligned} & \because \circ \\ & \hline 6 \\ & \hline \end{aligned}$ | $\begin{array}{l\|} \hline \varepsilon_{0}, \\ 0 \\ 0 \end{array}$ |  | $\bullet \bullet \bullet$ |  |  |  |  |  |  | $\begin{array}{\|c\|} \hline \bullet \bullet \\ \vdots \\ \vdots \\ \hline 0.0 \\ \hline \end{array}$ | -é |  |  | $\because \%$ |
| xxxx 0100 | $\left.\begin{gathered} \text { CG } \\ \text { RAM } \\ (1) \end{gathered} \right\rvert\,$ |  |  |  |  | : | $\%$ |  |  | $c_{0}^{\bullet \infty}$ |  | $\mid-0.0 \bullet$ | \&ob |  |  | 88 |
| xxxx 0101 | $\begin{array}{\|c\|} \hline \text { CG } \\ \text { RAM } \\ (2) \\ \hdashline \end{array}$ | $\begin{array}{\|c\|c\|} \hline \bullet \bullet & \theta^{\circ} \\ \hline \end{array}$ |  |  | $0$ | ${ }^{\bullet \bullet 0}$ | $$ | $\overbrace{}^{\circ \bullet}$ |  |  | 8 | $\bullet!$ | -••••••• | $\left\|\begin{array}{c} 000 \ell_{0}^{0} \\ 0.0 \end{array}\right\|$ |  | - 0 |
| xxxx 0110 | $\begin{gathered} \text { CG } \\ \text { RAM } \\ (3) \end{gathered}$ |  |  |  |  |  | $\%_{8}^{\circ}$ |  |  |  | $\mid$ |  |  | -000. |  |  |
| xxxx 0111 | $\begin{array}{\|c\|} \hline \text { CG } \\ \text { RAM } \\ \hline(4) \\ \hdashline--\mid \end{array}$ |  | -i | $\vdots$ | $\mid$ |  |  |  |  |  |  |  |  |  | $0$ |  |
| xxxx 1000 | $\left.\begin{gathered} \text { CG } \\ \text { RAM } \\ 1(1) \end{gathered} \right\rvert\,$ | $\left\|\begin{array}{c\|c} \bullet 0000 \\ \bullet \bullet 006 \end{array}\right\|$ | $\left\lvert\, \begin{gathered} \bullet^{\circ} \\ \vdots \\ \hdashline \bullet \end{gathered}\right.$ |  |  |  | 8 |  |  |  |  |  |  |  |  |  |
| xxxx 1001 | $\begin{gathered} \text { CG } \\ \text { RAM } \\ (2) \end{gathered}$ | $\left\lvert\, \begin{array}{ll} -0000 \\ 0000 \\ \vdots \end{array}\right.$ |  |  |  |  | ®ọ | $000$ | $8$ |  |  | \% |  |  | - | \% |
| xxxx 1010 | $\left.\begin{gathered} \text { CG } \\ \text { RAM } \\ -(3) \end{gathered} \right\rvert\,$ | $\stackrel{\bullet \bullet}{\bullet \bullet} \bullet$ | -eio | $\begin{aligned} & \text { :8 } \\ & \text { :8 } \end{aligned}$ | $\begin{gathered} \bullet \bullet \bullet \\ \bullet \\ \vdots \\ \vdots \\ \hline \end{gathered}$ | \|lot | $\bullet \text { •® }$ | $\left\|\begin{array}{c\|c} \bullet \because \bullet \bullet \\ \bullet \bullet \circ \end{array}\right\|$ | $\mid$ | ! |  |  |  |  |  | -0.0. |
| xxxx 1011 | $\begin{gathered} \text { CG } \\ \text { RAM } \\ (4) \end{gathered}$ | - ${ }^{\circ}$ | -o! | $\begin{aligned} & \text { :\& } \\ & \text { i8 } \end{aligned}$ | $\left[\begin{array}{c} 0 \circ \\ 00^{\circ} \\ \hline 0 \end{array}\right.$ |  |  | $!$ |  | \&o |  |  |  |  |  | -..0\% |
| xxxx 1100 | $\begin{array}{\|c\|} \hline \text { CG } \\ \text { RAM } \\ \text { (1) } \\ \hline \end{array}$ | Oesest | : |  | 2000 |  | -8. | : | $\begin{aligned} & \circ \circ \\ & \circ 0_{0} \\ & 0 \end{aligned}$ |  | : os |  |  |  |  | \% \% |
| xxxx 1101 | $\begin{array}{\|c\|} \hline \text { CG } \\ \text { RAM } \\ (2) \\ \hline 2) \end{array}$ | ! | 00000 | 100000 |  | $\begin{array}{r} \bullet 00 \\ 0 \\ \hline \end{array}$ |  |  |  |  |  |  |  |  | $\%_{8}^{\circ 000}$ | $\bullet$ |
| xxxx 1110 | $\begin{array}{\|c\|} \hline \text { CG } \\ \text { RAM } \\ (3) \\ \hdashline- \end{array}$ | $\|\because \because\|$ | 88 |  |  |  |  | -00\%. |  | on en | $\begin{array}{\|c\|c\|c\|} \hline 0008 \\ \hline 0006 \\ \hline \end{array}$ |  |  |  |  |  |
| xxxx 1111 | $\begin{array}{\|c\|} \hline \text { CG } \\ \text { RAM } \\ (4) \\ \hline \end{array}$ |  |  |  | $\begin{array}{\|l\|l\|} \hline \bullet \bullet 0 \\ \vdots & \vdots \\ \vdots \end{array}$ | ${ }^{00000}$ | $\vdots_{000}^{\circ 00}$ |  |  |  | $8 \overbrace{}^{\circ}$ |  |  | $80$ | $0_{0}^{\infty} 08$ | \% |

## Modifying Character Patterns

- Character pattern development procedure

The following operations correspond to the numbers listed in Figure 3:
a. Determine the correspondence between character codes and character patterns.
b. Create a listing indicating the correspondence between EPROM addresses and data.
c. Program the character patterns into an EPROM.
d. Send the EPROM to Hitachi.
e. Computer processing of the EPROM is performed at Hitachi to create a character pattern listing, which is sent to the user.
f. If there are no problems within the character pattern listing, a trial LSI is created at Hitachi and samples are sent to the user for evaluation. When it is confirmed by the user that the character patterns are correctly written, mass production of the LSI will proceed at Hitachi.

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Figure 3 Character Pattern Development Procedure

## Programming Character Patterns

This section explains the correspondence between addresses and data used to program character patterns in EPROM.

- Programming to EPROM

The HD66717 character generator ROM can generate $2405 \times 8$-dot character patterns. Table 7 shows correspondence between the EPROM address data and the character pattern.

## Handling Unused Character Patterns

1. EPROM data outside the character pattern area: This is ignored by the character generator ROM for display operation so any data is acceptable.
2. EPROM data in CGRAM area: Always fill with zeros.
3. Treatment of unused user patterns in the HD66717 EPROM: According to the user application, these are handled in either of two ways:
a. When unused character patterns are not programmed: If an unused character code is written into DDRAM, all its dots are lit, because the EPROM is filled with 1 s after it is erased.
b. When unused character patterns are programmed as 0s: Nothing is displayed even if unused character codes are written into DDRAM. (This is equivalent to a space.)

## Table 7 Example of Correspondence between EPROM Address Data and Character Pattern ( $5 \times 8$ Dots)



Notes: 1. EPROM addresses A11 to A4 correspond to a character code.
2. EPROM addresses $A 2$ to A0 specify the line position of the character pattern. EPROM address A3 should be set to 0 .
3. EPROM data O 4 to O 0 correspond to character pattern data.
4. Areas which are lit (indicated by shading) are stored as 1 , and unlit areas as 0 .
5. The eighth raster-row is also stored in the CGROM, and should also be programmed. If the eighth raster-row is used for a cursor, this data should all be set to zero.
6. EPROM data bits O 7 to O 5 are invalid. 0 should be written in all bits.

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Table 8 Example of Relationships between Character Code (DDRAM) and Character Pattern (CGRAM Data)


Notes: 1. The lower 2 bits of the character code correspond to the upper two bits of the CGRAM address (2 bits: 4 types).
2. CGRAM address bits 0 to 2 designate the character pattern raster-row position. The 8th rasterrow is the cursor position and its display is formed by a logical OR with the cursor.
3. The upper three bits of the CGRAM data are invalid; use the lower five bits.
4. When the upper four bits (bits 7 to 4 ) of the character code are 0 , CGRAM is selected.

Bits 3 and 2 of the character code are invalid (*). Therefore, for example, the character codes (00)H and (08)H correspond to the same CGRAM address.
5. A set bit in the CGRAM data corresponds to display selection, and 0 to non-selection.

* Indicates no effect.

Table 9 Correspondence between Segment Display SEGRAM Addresses (ASEG) and Driver Signals

| ASEG Address |  |  |  | Segment Signals |  |  |  |  |  |  |  | Common Signal |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MSB |  |  | LSB | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |  |
| 1 | 0 | 0 | 0 | * | * | * | SEG1, SEG21, SEG41 | SEG2, SEG22, SEG42 | SEG3, SEG23, SEG43 | SEG4, SEG24, SEG44 | SEG5, SEG25, SEG45 | COMS1 |
| 1 | 0 | 0 | 1 | * | * | * | SEG6, SEG26, SEG46 | $\begin{aligned} & \hline \text { SEG7, } \\ & \text { SEG27, } \\ & \text { SEG47 } \end{aligned}$ | SEG8, SEG28, SEG48 | $\begin{aligned} & \text { SEG9, } \\ & \text { SEG29, } \\ & \text { SEG49 } \end{aligned}$ | SEG10, SEG30, SEG50 | COMS1 |
| 1 | 0 | 1 | 0 | * | * | * | SEG11, SEG31, SEG51 | $\begin{aligned} & \text { SEG12, } \\ & \text { SEG32, } \\ & \text { SEG52 } \end{aligned}$ | SEG13, SEG33, SEG53 |  | SEG15, SEG35, SEG55 | COMS1 |
| 1 | 0 | 1 | 1 | * | * | * | SEG16, SEG36, SEG56 | SEG17 SEG37, SEG57 | SEG18, SEG38, SEG58 | SEG19, SEG39, SEG59 | SEG20, SEG40, SEG60 | COMS1 |
| 1 | 1 | 0 | 0 | * | * | * | $\begin{aligned} & \hline \text { SEG1, } \\ & \text { SEG21, } \\ & \text { SEG41, } \end{aligned}$ | SEG2, SEG22, <br> SEG42 | SEG3, SEG23, SEG43 | SEG4, SEG24, SEG44 | SEG5, SEG25, SEG45 | COMS2 |
| 1 | 1 | 0 | 1 | * | * | * | SEG6, SEG26, SEG46 | SEG7, SEG27, SEG47 | SEG8, SEG28, SEG48 | SEG9, SEG29, SEG49 | SEG10, SEG30, SEG50 | COMS2 |
| 1 | 1 | 1 | 0 | * | * | * | SEG11, SEG31, SEG51 | SEG12, SEG32, SEG52 | SEG13, SEG33, SEG53 | SEG14, SEG34, SEG54 | SEG15, SEG35, SEG55 | COMS2 |
| 1 | 1 | 1 | 1 | * | * | * | SEG16, SEG36, SEG56 | SEG17 SEG37, SEG57 | SEG18, SEG38, SEG58 | SEG19, SEG39, SEG59 | SEG20, SEG40, SEG60 | COMS2 |

Notes: 1. When the SFT pin is grounded, the SEG1 pin output is connected to the far left of the LCD panel, and when the SFT pin is high, the SEG60 pin output is connected to the far left.
2. SEG1 to SEG20 data is identical to SEG21 to SEG40 and SEG41 to SEG60 data.
3. The lower five bits ( D 4 to D 0 ) of SEGRAM data determine on or off display of each segment. A segment is selected (turned on) when the corresponding data is 1 , and is deselected (turned off) when the corresponding data is 0 . The upper three bits (D7 to D5) are invalid.

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Table 10 Correspondence between Annunciator Display Addresses (AAN) and Driver Signals

| AAN Address |  |  |  | Annunciator Segment Signals |  |  |  |  |  |  |  | Common Signal |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MSB |  |  | LSB | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |  |
| 0 | 0 | 0 | 0 | ASEG1 | ASEG1 | ASEG2 | ASEG2 | ASEG3 | ASEG3 | ASEG4 | ASEG4 | ACOM |
|  |  |  |  | Blink | Data | Blink | Data | Blink | Data | Blink | Data |  |
| 0 | 0 | 0 | 1 | ASEG5 | ASEG5 | ASEG6 | ASEG6 | ASEG7 | ASEG7 | ASEG8 | ASEG8 | ACOM |
|  |  |  |  | Blink | Data | Blink | Data | Blink | Data | Blink | Data |  |
| 0 | 0 | 1 | 0 | ASEG9 | ASEG9 | ASEG10 | ASEG10 | - | - | - | - | ACOM |
|  |  |  |  | Blink | Data | Blink | Data | * | * | * | * |  |

Notes: 1. The annunciator is turned on when the corresponding even bit (data) is 1 , and is turned off when 0.
2. The turned-on annunciator blinks when the corresponding odd bit (blink) is 1 . Blinking is provided by repeatedly turning on the annunciator for 32 frames and then turning it off for the next 32 frames.

## Instructions

## Outline

Only the instruction register (IR) and the data register (DR) of the HD66717 can be controlled by the MPU. Before starting internal operation of the HD66717, control information is temporarily stored in these registers to allow interfacing with various peripheral control devices or MPUs which operate at different speeds. The internal operation of the HD66717 is determined by signals sent from the MPU. These signals, which include register selection (RS), read/write (R/W), and the data bus (DB0 to DB7), make up the HD66717 instructions (Table 17). There are four categories of instructions that:

- Control display
- Control power management
- Set internal RAM addresses
- Perform data transfer with internal RAM

Normally, instructions that perform data transfer with internal RAM are used the most. However, autoincrementation by 1 (or auto-decrementation by 1) of internal HD66717 RAM addresses after each data write can lighten the program load of the MPU.

While an instruction is being executed for internal operation, or during reset, no instruction other than the busy flag/address read instruction can be executed.

Because the busy flag is set to 1 while an instruction is being executed, check it to make sure it is 0 before sending another instruction from the MPU. If an instruction is sent without checking the busy flag, the time between the first instruction issue and next instruction issue must be longer than the instruction execution time itself. Refer to Table 16 for the list of each instruction execution cycles (clock pulses). The execution time depends on the operating clock frequency (oscillation frequency).

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## Instruction Description

## Status Read

The status read instruction (Figure 4) reads the busy flag (BF) indicating that the system is now internally operating on a previously received instruction. If BF is 1 , the internal operation is in progress. The next instruction will not be accepted until BF is reset to 0 . Check the BF status before the next write operation. At the same time, the value of the address counter in binary AAAAAAA is read out. This address counter is used by both CGRAM, DDRAM, and SEGRAM addresses, and its value is determined by the previous instruction.

## Clear Display

The clear display instruction (Figure 5) writes space code (20)H (character pattern for character code (20)H must be a blank pattern) into all DDRAM addresses. It then sets DDRAM address 0 into the address counter. It also sets I/D to 1 (increment mode) in entry mode.
RS R/W DB7

| 0 | 1 | BF | A | A | A | A | A | A | A |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Figure 4 Status Read Instruction

| RS R/W DB7 ------------------------ DB0 |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |

Figure 5 Clear Display Instruction

## Return Home

The return home instruction (Figure 6) sets DDRAM address 0 into the address counter. The DDRAM contents do not change. The cursor or blinking goes to the top left of the display.

## Start Oscillator

The start oscillator instruction (Figure 7) re-starts the oscillator from a halt state in standby mode. After issuing this instruction, wait at least 10 ms for oscillation to become stable before issuing the next instruction. (Refer to the Standby Mode section.)

## Entry Mode

The entry mode instruction (Figure 8) includes the I/D and OSC bits.
I/D: Increments $(\mathrm{I} / \mathrm{D}=1)$ or decrements $(\mathrm{I} / \mathrm{D}=0)$ the DDRAM address by 1 when a character code is written into or read from DDRAM. The cursor or blinking moves to the right when incremented by 1 and to the left when decremented by 1 . The same applies to writing and reading of CGRAM and SEGRAM.

OSC: Divides the external clock frequency by four ( $\mathrm{OSC}=1$ ) using the resulting clock as an internal operating clock. The execution time for this instruction and subsequent ones is therefore quadrupled. The execution time of clearing this bit $(\mathrm{OSC}=0)$ is also quadrupled. (For application of this instruction, refer to the Partial-Display-Off Function section.)

|  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 |

Figure 6 Return Home Instruction

| RS R/W DB7 ------------------------ DB0 |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 |

Figure 7 Start Oscillator Instruction

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## Cursor Control

The cursor control (Figure 9) includes the B/W, C, and B bits.
B/W: When B/W is 1 , the character at the cursor position is cyclically (every 32 frames) displayed with black-white inversion.
$\mathbf{C}$ : The cursor is displayed on the 8 th raster-row when C is 1 . The cursor is displayed using 5 dots in the 8 th raster-row for $5 \times 8$-dot character font.

B: The character indicated by the cursor blinks when B is 1 . The blinking is displayed as switching between all black dots and displayed characters every 32 frames. The cursor and blinking can be set to display simultaneously. When LC and $\mathrm{B}=1$, the blinking is displayed as switching between all white dots and displayed characters.

Figure 10 shows cursor control examples.
RS R/W DB7

| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | I/ | OSC |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Figure 8 Entry Mode Instruction
RS R/W DB7

| 0 | 0 | 0 | 0 | 0 | 0 | 1 | $B / W$ | $C$ | $B$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Figure 9 Cursor Control Instruction

## Display On/Off Control

The display on/off control instruction (Figure 11) includes DC, DS, and LC bits.
DC: The character display is on when DC is 1 and off when DC is 0 . When off, the display data remains in DDRAM, and can be displayed instantly by setting DC to 1 .

DS: When $\mathrm{DS}=1$, segment display for icons and marks that is controlled by the multiplexing drive method is turned on and when $\mathrm{DS}=0$, it is turned off.

When both DC and $\mathrm{DS}=0$, multiplexing drive is halted, setting the outputs from SEG1 to SEG60, COM1 to COM32, and COMS1 and COMS2 to $\mathrm{V}_{\mathrm{CC}}$ level to turn off the display. This can suppress current for LCD charging or discharging due to LCD driving operations.

LC: When $\mathrm{LC}=1$, a cursor attribute is assigned to the line that contains the address counter (AC) value. Cursor mode can be selected with the B/W, C, and B bits. Refer to the Line-Cursor Display section.


Figure 10 Cursor Control Examples
RS R/W DB7

| 0 | 0 | 0 | 0 | 0 |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Figure 11 Display On/Off Instruction

## HD66717

## Power Control

The cursor control instruction (Figure 12) includes the AMP, SLP, and STB bits.
AMP: When AMP = 1, each voltage-follower for V1 to V5 pins and the booster are turned on. When AMP $=0$, current consumption can be reduced while character or segment display controlled by the multiplexing drive method is not being used.

SLP: When SLP $=1$, the HD66717 enters sleep mode, where all the internal operations are halted except for annunciator display function and the R-C oscillator, thus reducing current consumption. Refer to the Sleep Mode section. Only the following instructions can be executed during sleep mode.

1. Annunciator address set (AAN)
2. Annunciator data write
3. Annunciator display on or off $(\mathrm{DA}=1$ or 0$)$
4. Voltage-follower on or off $(\mathrm{AMP}=1$ or 0$)$
5. Standby mode set $(\mathrm{STB}=1)$
6. Sleep mode cancel $(\mathrm{SLP}=0)$

During sleep mode, other RAM data and instructions cannot be updated but they are retained.
STB: When STB $=1$, the HD66717 enters standby mode, where the device completely stops, halting all the internal operations including the internal R-C oscillator and no external clock pulses are supplied. However, annunciator display alone is available when the alternating signal for annunciator-driving signals is supplied to the EXM pin. When the annunciator display is not needed, make sure to turn off display ( $\mathrm{DA}=0$ ). Refer to the Standby Mode section. Only the following instructions can be executed during standby mode.

1. Annunciator address set (AAN)
2. Annunciator data write
3. Annunciator display on or off $(\mathrm{DA}=1$ or 0$)$
4. Voltage-follower on or off $(\mathrm{AMP}=1$ or 0$)$
5. Start oscillator
6. Standby mode cancel $(\mathrm{STB}=0)$

During standby mode, RAM data and other instructions may be lost; they must be set again after standby mode is cancelled.
RS R/W DB7

| 0 | 0 | 0 | 0 | 0 | 1 | 1 | AMP | SLP | STB |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Figure 12 Power Control Instruction

## Display Control

The display control instruction (Figure 13) includes the NL and DL bits.
NL1, NL0: Designates the number of display lines. This value determines the LCD drive multiplexing duty ratio (Table 11). The address assignment is the same for all display line modes.

DL3-DL1: Doubles the height of characters on a specified line. The first, second, or third line is doubled in height when DL1, DL2, or DL3 = 1, respectively. Two lines can be simultaneously doubled in a 4-line display. Refer to the Double-Height Display section.

| RS R/W DB7 .------------------------- DB0 |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 |  |  |  | LO | DL3 | DL2 | 2 DL1 |

Figure 13 Display Control Instruction
Table 11 NL Bits and Display Lines

| NL1 | NLO | Number of Display Lines | LCD Drive Multiplexing Duty Ratio |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 1 | $1 / 10$ |
| 0 | 1 | 2 | $1 / 18$ |
| 1 | 0 | 3 | $1 / 26$ |
| 1 | 1 | 4 | $1 / 34$ |

## HD66717

## Contrast Control

The contrast control instruction (Figure 14) includes the SN and CT bits.
SN2: Combined with the SN1 and SN0 bits described in the Scroll Control section to select the first line to be scrolled (display-start line).

CT3-CT0: Controls the LCD drive voltage (potential difference between $\mathrm{V}_{\mathrm{CC}}$ and V5) to adjust contrast (Figure 15 and Table 12). Refer to the Contrast Adjuster section.

$$
\begin{aligned}
& \text { RS R/W DB7 } \\
& \begin{array}{|l|l|l|l|l|l|l|}
\hline 0 & 0 & 0 & 1 & 0 & \text { SN2 } & \text { CT3 } \\
\hline
\end{array}
\end{aligned}
$$

## Figure 14 Contrast Control Instruction



Figure 15 Contrast Adjuster

## Scroll Control

The scroll control instruction (Figure 16) includes the SN and SL bits.
SN1, SN0: Combined with the SN2 bit described in the Contrast Control section to select the top line to be displayed (display-start line) through the data output from the COM1 pin (Table 13). After first five lines are displayed from the top line, the cycle is repeated and scrolling continues.

Table 12 CT Bits and Variable Resistor Value of Contrast Adjuster

| CT3 | CT2 | CT1 | CT0 | Variable Resistor Value (VR) |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | $6.4 \times R$ |
| 0 | 0 | 0 | 1 | $6.0 \times \mathrm{R}$ |
| 0 | 0 | 1 | 0 | $5.6 \times \mathrm{R}$ |
| 0 | 0 | 1 | 1 | $5.2 \times \mathrm{R}$ |
| 0 | 1 | 0 | 0 | $4.8 \times \mathrm{R}$ |
| 0 | 1 | 0 | 1 | $4.4 \times \mathrm{R}$ |
| 0 | 1 | 1 | 0 | $4.0 \times \mathrm{R}$ |
| 0 | 1 | 1 | 1 | $3.6 \times \mathrm{R}$ |
| 1 | 0 | 0 | 0 | $3.2 \times \mathrm{R}$ |
| 1 | 0 | 0 | 1 | $2.8 \times \mathrm{R}$ |
| 1 | 0 | 1 | 0 | $2.4 \times \mathrm{R}$ |
| 1 | 0 | 1 | 1 | $2.0 \times \mathrm{R}$ |
| 1 | 1 | 0 | 0 | $1.6 \times \mathrm{R}$ |
| 1 | 1 | 0 | 1 | $1.2 \times \mathrm{R}$ |
| 1 | 1 | 1 | 0 | $0.8 \times \mathrm{R}$ |
| 1 | 1 | 1 | 1 | $0.4 \times \mathrm{R}$ |



Figure 16 Scroll Control Instruction

## HD66717

SL2-SL0: Selects the top raster-row to be displayed (display-start raster-row) in the display-start line specified by SN2 to SN0. Any raster-row from the first to eighth can be selected (Table 14). This function is used to perform vertical smooth scroll together with SN2 to SN0. Refer to the Vertical Smooth Scroll section.

Table 13 SN Bits and Display-Start Lines

| SN2 | SN1 | SNO | Display-Start Line |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 1st line |
| 0 | 0 | 1 | 2nd line |
| 0 | 1 | 0 | 3rd line |
| 0 | 1 | 1 | 4th line |
| 1 | $0 / 1$ | $0 / 1$ | 5th line |

Table 14 SN Bits and Display-Start Raster-Rows

| SL2 | SN1 | SLO | Display-Start Raster-Row |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 1st raster-row |
| 0 | 0 | 1 | 2nd raster-row |
| 0 | 1 | 0 | 3rd raster-row |
| 0 | 1 | 1 | 4th raster-row |
| 1 | 0 | 0 | 5th raster-row |
| 1 | 0 | 1 | 6th raster-row |
| 1 | 1 | 0 | 7th raster-row |
| 1 | 1 | 1 | 8th raster-row |

## Annunciator/SEGRAM Address Set

The annunciator/SEGRAM address set instruction (Figure 17) includes the DA and A bits.
DA: Turns annunciator display on or off. When $\mathrm{DA}=1$, annunciator display is turned on and driven statically. When $\mathrm{DA}=0$, annunciator display is turned off with ASEG1 to ASEG10 and ACOM pins held to $\mathrm{V}_{\mathrm{CC}}$ level.

The internal operating clock supply is halted during standby mode; make sure to turn off display ( $\mathrm{DA}=$ 0 ) if the external alternating signal is not supplied. Refer to the Segment Display and Annunciator Display section and the Standby Mode section.

AAAA: Used for setting the SEGRAM address in the address counter (AC) or for setting an annunciator address. The SEGRAM addresses range from 1000 H to 1111 H ( 8 addresses), while the annunciator addresses range from 0000 H to 0010 H ( 3 addresses).

The annunciator address is directly set without using the address counter, and consequently must be updated for each access. The annunciator address can be set even during sleep and standby modes.

Once the SEGRAM address is set, data in the SEGRAM can be accessed consecutively since the address counter is automatically incremented or decremented by one according to the I/D bit setting after each access. The SEGRAM address cannot be set during sleep or standby mode.
RS R/W DB7

| 0 | 0 | 1 | 0 | 0 | DA | A | A | A | A |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Figure 17 Annunciator/SEGRAM Address Set Instruction

## HD66717

## CGRAM Address Set

The CGRAM address set instruction (Figure 18) includes the A bits.
AAAAA: Used for setting the CGRAM address in the address counter (AC). The CGRAM addresses range from 00 H to 1 FH ( 32 addresses) (Table 15).

Once the CGRAM address is set, data in the CGRAM can be accessed consecutively since the address counter is automatically incremented or decremented according to the I/D bit setting after each access. The CGRAM address cannot be set during sleep or standby mode.
RS R/W DB7

| 0 | 0 | 1 | 0 | 0 | 1 |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Figure 18 CGRAM Address Set Instruction
Table 15 CGRAM Addresses and Character Codes

| Displayed Character | CGRAM Address | Character Codes |
| :--- | :--- | :--- |
| 1st character | 00 H to 07 H | 00 H |
| 2nd character | 08 H to 0 FH | 01 H |
| 3rd character | 10 H to 17 H | 02 H |
| 4th character | 18 H to 1 FH | 03 H |

## DDRAM Address Set

The DDRAM address set instruction (Figure 19) includes the A bits.
AAAAAAA: Used for setting the DDRAM address in the address counter (AC). The DDRAM addresses range from 00 H to 4 BH ( 60 addresses) (Table 16).

Once the DDRAM address is set, data in the DDRAM can be accessed consecutively since the address counter is automatically incremented or decremented according to the I/D bit setting after each access. Here, invalid addresses are automatically skipped. The DDRAM address cannot be set during sleep or standby mode.
RS R/W DB7

| 0 | 0 | 1 | 1 | 0 | 0 | 0 | 0 | A | A |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 0 | Upper bits |  |  |  |  |  |  |  |  |
| 0 | 0 | 1 | 1 | 1 | A | A | A | A | A |

Figure 19 DDRAM Address Set Instruction

Table 16 DDRAM Addresses and Invalid Addresses

| Displayed Line | DDRAM Address | Invalid Addresses |
| :--- | :--- | :--- |
| 1st line | 00 H to 0 BH | 0 CH to 0 FH |
| 2nd line | 10 H to 1 BH | 1 CH to 1 FH |
| 3rd line | 20 H to 2 BH | 2 CH to 2 FH |
| 4th line | 30 H to 3 BH | 3 CH to 3 FH |
| 5th line | 40 H to 4 BH | 4 CH and subsequent addresses |

## HD66717

## Write Data to RAM

The write data to RAM instruction (Figure 20) writes 8-bit data to annunciator or DDRAM, or lower 5-bit data to SEGRAM or CGRAM that is selected by the previous specification of the address set instruction (annunciator/SEGRAM address set, CGRAM address set, or DDRAM address set).

After a write, the address is automatically incremented or decremented by 1 according to the $\mathrm{I} / \mathrm{D}$ bit setting in the entry mode instruction.

The annunciator address is not automatically updated; it must be specifically updated to write data to a different address. During sleep and standby modes, DDRAM, CGRAM, or SEGRAM cannot be accessed.

## Read Data from RAM

The read data from RAM instruction (Figure 21), reads 8-bit data from DDRAM, or 5-bit binary data from CGRAM or SEGRAM that is selected by the previous specification of the address set instruction (SEGRAM address set, CGRAM address set, or DDRAM address set). The unused upper three bits of CGRAM or SEGRAM data are read as 000; annunciator data cannot be read. If no address is specified by the address set instruction just before this instruction, the first data read will be invalid. When executing serial read instructions, the next address is normally read from the next address.

After a read, the address is automatically incremented or decremented by 1 according to the I/D bit setting in the entry mode instruction.

Table 17 lists the above instructions.
RS R/W DB7

| 1 | 0 | D | D | D | D | D | D | D | D |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Figure 20 Write Data to RAM Instruction
RS R/W DB7

| 1 | 1 | D | D | D | D | D | D | D | D |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Figure 21 Read Data from RAM Instruction

Table 17 Instruction List

| Instruction | No. | Code |  |  |  |  |  |  |  |  |  | Description | Execution Cycle *1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | R/W | RS | DB7 | DB6 | DB5 | DB4 | DB3 | DB2 | DB1 | DB0 |  |  |
| Status | SR | 1 | 0 | BF | AC | AC | AC | AC | AC | AC | AC | Reads busy flag (BF), which indicates internal operations are being performed, and reads address counter (AC). | 0 |
| Clear display | CL | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | Clears entire display and sets DDRAM address 0 in address counter. | 310 |
| Return home | CH | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | Sets DDRAM address 0 in address counter. | 10 |
| Start oscillator | OS | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | Starts oscillation during standby mode. | - |
| Entry mode set | EM | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | I/D | OSC | Sets address update direction after RAM access (I/D), and system clock division (OSC). | 10 |
| Cursor control | CR | 0 | 0 | 0 | 0 | 0 | 0 | 1 | B/W | C | B | Sets black-white inverting cursor (B/W), 8th rasterrow cursor (C), and blink cursor (B). | 10 |
| Display on/off control | DO | 0 | 0 | 0 | 0 | 0 | 1 | 0 | DC | DS | LC | Sets character display on/off (DC), segment display on/off (DS), and line-cursor on/off (LC). | 10 |
| Power control | PW | 0 | 0 | 0 | 0 | 0 | 1 | 1 | AMP | SLP | STB | Turns on voltage-follower and booster (AMP), and sets sleep mode (SLP) and standby mode (STB). | 10 |
| Display control | DC | 0 | 0 | 0 | 0 | 1 | NL1 | NLO | DL3 | DL2 | DL1 | Sets the number of display lines (NL) and the line to be doubled in height. | 10 |
| Contrast control | CN | 0 | 0 | 0 | 1 | 0 | SN2 | CT3 | CT2 | CT1 | CTO | Sets the display-start line (SN2) and contrastadjusting value (CT). | 10 |
| Scroll control | SC | 0 | 0 | 0 | 1 | 1 | SN1 | SN0 | SL2 | SL1 | SLO | Sets the display-start line (SN) and display-start raster-row (SL). | 10 |
| Annunciator /SEGRAM address set | AS | 0 | 0 | 1 | 0 | 0 | DA | AAN/ <br> $\mathrm{A}_{\text {seG3 }}$ | AAN/ $\mathrm{A}_{\mathrm{sEG} 2}$ | AAN/ <br> $\mathrm{A}_{\text {sEG1 }}$ | AAN/ <br> $\mathrm{A}_{\text {sego }}$ | Turns on annunciator display and sets annunciator/SEGRAM address. | 10 |
| CGRAM address set | CA | 0 | 0 | 1 | 0 | 1 | $\mathrm{A}_{\text {cG4 }}$ | $\mathrm{A}_{\text {cG3 }}$ | $\mathrm{A}_{\text {cG2 }}$ | $\mathrm{A}_{\text {cG1 }}$ | $\mathrm{A}_{\text {cGo }}$ | Sets the initial CGRAM address to the address counter. | 10 |
| DDRAM address set (upper bits) | DA | 0 | 0 | 1 | 1 | 0 | 0 | 0 | 0 | $\mathrm{A}_{\text {DD6 }}$ | $\mathrm{A}_{\text {DD5 }}$ | Sets the initial higher DDRAM address to the address counter. | 10 |
| DDRAM address set (lower bits) | DA | 0 | 0 | 1 | 1 | 1 | $\mathrm{A}_{\text {DD } 4}$ | $\mathrm{A}_{\text {D } 3}$ | $\mathrm{A}_{\text {DD2 }}$ | $\mathrm{A}_{\mathrm{DD} 1}$ | $\mathrm{A}_{\text {DD }}$ | Sets the initial lower DDRAM address to the address counter. | 10 |

## HD66717

Table 17 Instruction List (cont)


## Reset Function

## Initializing by Internal Reset Circuit

The HD66717 is internally initialized by RESET input. During reset, the system executes the instructions as described below. Here, the busy flag ( BF ) therefore indicates a busy state ( $\mathrm{BF}=1$ ), accepting no instruction or RAM data access from the MPU. Here, reset input must be held at least 10 ms .

After releasing power-on reset, clear display instruction is operated, so wait for 1,000 clock-cycles or more.

Make sure to reset the HD66717 immediately after power-on reset in $\mathrm{I}^{2} \mathrm{C}$ bus mode.

1. Instruction set initialization
a. Clear display:

Writes 20H to DDRAM after releasing reset
b. Return home

Sets the address counter (AC) to 00 H to select the DDRAM
c. Start oscillator
d. Entry mode

I/D = 1: Increment by 1
OSC $=0$ : Clock frequency not divided
e. Cursor control
$B / W=0$ : White-black inverting cursor off
$\mathrm{C}=0$ : 8th raster-row cursor off
B $=0$ : Blink cursor off
f. Display on/off control

DC $=0$ : Character display off
DS $=0$ : Segment display off
LC $=0$ : Line-cursor off
g. Power control

AMP $=0$ : LCD power supply off
SLP $=0$ : Sleep mode off
STB $=0$ : Standby mode off
h. Display control

NL1, NL0 $=11$ : 4-line display ( $1 / 34$ multiplexing duty ratio)
DL3-DL1 = 000: Double-height display off
i. Contrast adjust
$\mathrm{CT}=0000$ : Weak contrast
j. Scroll control

SN2-SN0 $=000$ : First line displayed at the top
SL2-SL0 $=000$ : First raster-row displayed at the top of the first line

## HD66717

k. Annunciator control

DA = 0: Annunciator display off
2. RAM data initialization
a. DDRAM

All addresses are initialized to 20 H by the clear display instruction
b. CGRAM/SEGRAM

Not automatically initialized by reset input; must be initialized by software while display is off ( DC and $\mathrm{DS}=0$ )
c. Annunciator data

Not automatically initialized by reset input; must be initialized by software while display is off ( $\mathrm{DA}=0$ )
3. Output pin initialization
a. LCD driver output pins (SEG/COM, ASEG/ACOM): Outputs $\mathrm{V}_{\mathrm{CC}}$ level
b. Booster output pins (V5OUT2 and V5OUT3): Outputs GND level
c. Oscillator output pin (OSC2): Outputs oscillation signal

## Transferring Serial Data

## $I^{2} \mathbf{C}$ Bus Interface

Grounding the IM1 and IM0 pins (interface mode pins) allows serial data transfer conforming to the $\mathrm{I}^{2} \mathrm{C}$ bus interface over the serial data line (SDA) and serial transfer clock line (SCL). Here, the HD66717 operates in an exclusive-receive slave mode.

The HD66717 initiates serial data transfer by transferring the first byte when a high SCL level at the falling edge of the SDA input is sampled; it ends serial data transfer when a high SCL level at the rising edge of the SDA input is sampled.

The HD66717 is selected when the higher six bits of the 7-bit slave address in the first byte transferred from the master device match the 6-bit device identification code assigned to the HD66717. The HD66717, when selected, receives the subsequent data strings. Any identification code can be assigned by the DB5/ID5 to DB0/ID0 pins; select an appropriate code that is not assigned to any other slave device. The higher four bits (ID5 to ID2) of this identification code is recommended as 0111. Two different slave addresses must be assigned to a single HD66717 because the least significant bit (LSB) of the slave address is used as a register select bit (RS): when $\mathrm{RS}=0$, an instruction can be issued and when $\mathrm{RS}=1$, data can be written to a RAM. The eighth bit of the first byte (R/W bit) must be 0 since the HD66717 exclusively receives data.

The ninth bit of the first byte is a receive-data acknowledge bit (ACK). When the received slave address matches the device ID code, the HD66717 pulls down the ACK bit to a low level. Therefore, the ACK output buffer is an open-drain structure, only allowing low-level output. However, the ACK bit is undetermined immediately after power-on; make sure to initialize the LSI using the RESET* input.

After identifying the address in the first byte, the HD66717 receives the subsequent data as an HD66717 instruction or as RAM data. Having received 8-bit data normally, the HD66717 pulls down the ninth bit (ACK) to a low level. Therefore, if the ACK is not returned, the data must be transferred again. Multiple bytes of data can be consecutively transferred until the transfer-end condition is satisfied. Here, when the serial data transfer rate is longer than that of the HD66717 instruction execution cycle, effective data transfer is possible without retransmission (see Table 17, Instruction List). Note that the display-clear instruction alone requires longer execution time than the others.

Table 18 illustrates the first bytes of $\mathrm{I}^{2} \mathrm{C}$ bus interface data and Figure 22 shows the $\mathrm{I}^{2} \mathrm{C}$ bus interface timing sequence .

## HD66717

Table 18 First Bytes of $I^{2} \mathbf{C}$ Bus Interface Data
Transferred Bit String


b) Consecutive data transfer timing

Note: Transfer the instruction 2 ACK after instruction 1 has been executed.
Figure $22 I^{2} \mathbf{C}$ Bus Interface Timing Sequence

## Clock-Synchronized Serial Interface

Setting the IM1 and IM0 pins (interface mode pins) to the GND and high levels, respectively, allows standard clock-synchronized serial data transfer, using the chip select (CS*), SDA, and SCL lines. Here, the HD66717 exclusively receives data.

The HD66717 initiates serial data transfer by transferring the start byte at the falling edge of the CS* input. It ends serial data transfer at the rising edge of the CS* input.

The HD66717 is selected when the 6-bit chip address in the start byte transferred from the transmitting device matches the 6-bit (device) identification code assigned to the HD66717. The HD66717, when selected, receives the subsequent data strings. Any identification code can be assigned by the DB5/ID5 to DB0/ID0 pins. Two different chip addresses must be assigned to a single HD66717 because the seventh bit of the start byte is used as a register select bit (RS): when $\mathrm{RS}=0$, an instruction can be issued and when $\mathrm{RS}=1$, data can be written to a RAM. The eighth bit of the start byte must be 0 .

After receiving the start byte, the HD66717 receives the subsequent data as an HD66717 instruction or as RAM data. Data is transferred with the MSB first. To transfer data consecutively, adjust the data transfer rate so that the HD66717 can complete the current instruction before the eighth bit of the next instruction is transferred. See Table 17, Instruction List. If the next instruction is received during execution of the previous instruction, the next instruction will be ignored. Note that the display-clear instruction alone requires longer execution time than the others.

Figure 23 shows the clock-synchronised serial interface timing sequence.

## HD66717


(a) Basic data transfer (receive) timing

(b) Consecutive data transfer timing

Note: Adjust the transfer rate so that the HD66717 can complete instruction 1 before the 8th bit of instruction 2 is transferred

Figure 23 Clock-Synchronized Serial Interface Timing Sequence

## Transferring Parallel Data

## Interface with an 8-Bit MPU

Eight-bit data can be transferred in parallel by setting the IM1 and IM0 pins to the $\mathrm{V}_{\mathrm{CC}}$ and GND levels, respectively (Figure 24). The HD66717 can interface directly with an 8-bit bus synchronized with the E clock, or with an 8 -bit MCU through an I/O port (Figure 25). When the number of I/O lines or chip packaging size is limited, a 4-bit bus interface or even serial data transfer should be used.


Figure 24 8-Bit Parallel Data Transfer Timing Sequence


Figure 25 8-Bit MPU Interface

## HD66717

## Interface with a 4-Bit MPU

Four-bit data can be transferred in parallel by setting both the IM1 and IM0 pins to the $\mathrm{V}_{\mathrm{CC}}$ level (Figure 26). Four-bit data representing higher or lower bits of 8 -bit instructions or 8 -bit RAM data can be transferred in that order.

The HD66717 can forcibly reset the counter that counts the number of higher and lower 4-bit data transfers in a 4-bit bus interface. This function, called transfer-syncronization, can be performed by writing a special instruction containing 0000 four consecutive times (Figure 27). For example, when a data transfer sequence becomes disordered due to noise or some undesired factor, this function resets the counter and thus enables resuming data transfer from the higher 4 bits. Using this function at specified intervals prevents display-system crash.


Figure 26 4-Bit Parallel Data Transfer Timing Sequence


Figure 27 4-Bit Data Transfer Synchronization

## Oscillator Circuit

The HD66717 can either be supplied with operating clock pulses externally (external clock mode) or oscillate using an internal R-C oscillator and an external oscillator-resistor (internal oscillation mode), as shown in Figure 28. An appropriate oscillator-resistor must be used to obtain the optimum clock frequency according to the number of display lines (Table 18). Instruction execution times change in proportion to the operating clock frequency or R-C oscillation frequency; MPU data transfer rate must be appropriately adjusted (see Table 17, Instruction List). Figure 29 shows a sample LCD drive output waveform, where 4 -lines are displayed with $1 / 34$ multiplexing duty ratio.


Figure 28 Oscillator Circuit
Table 19 Oscillation Frequency and LCD Frame Frequency

| Item |  | 1-Line Display NL1, NLO = 00 | 2-Line Display NL1, NLO = 01 | 3-Line Display NL1, NLO = 10 | 4-Line Display NL1, NLO = 11 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Multiplexing duty ratio |  | 1/10 | 1/18 | 1/26 | 1/34 |
| Oscillator resistance ( $\mathrm{R}_{\mathrm{t}}$ ) | $\mathrm{V}_{\mathrm{cc}}=3 \mathrm{~V}$ | $620 \mathrm{k} \Omega$ | $300 \mathrm{k} \Omega$ | $200 \mathrm{k} \Omega$ | $150 \mathrm{k} \Omega$ |
| CR oscillator frequency |  | 40 kHz | 85 kHz | 120 kHz | 160 kHz |
| Frame frequency |  | 67 Hz | 79 Hz | 77 Hz | 78 Hz |



Figure 29 LCD Drive Output Waveform Example (4-line display with 1/34 multiplexing duty ratio)

## Power Supply for Liquid Crystal Display Drive

## When External Power Supply and Internal Operational Amplifiers are Used

To supply LCD drive voltage directly from the external power supply without using the internal booster, circuits should be connected as shown in Figure 30. Here, contrast can be adjusted through the CT bits of the contrast-control instruction.

The HD66717 incorporates a voltage-follower operational amplifier for each of V1 to V5 to reduce current flowing through the internal bleeder-resistors, which generate different levels of liquid-crystal drive voltages. Thus, potential differences between $\mathrm{V}_{\mathrm{CC}}$ and V 1 and between $\mathrm{V}_{\mathrm{EE}}$ and V 5 must be 0.4 V or greater. Note that the OPOFF pin must be grounded when using the operational amplifiers.

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Note: 1. Potential differences between $\mathrm{V}_{\mathrm{CC}}$ and V 1 and between V 5 and $\mathrm{V}_{\mathrm{EE}}$ must be 0.4 V or greater, particularly for low-duty drive such as 1 -line display.
2. When the internal operational amplifiers cannot fully drive the LCD panel used, an appropriate capacitor must be inserted between each output of V1OUT to V5OUT and $\mathrm{V}_{\mathrm{CC}}$ to stabilize the operational amplifier output.

Figure 30 External Power Supply Circuit Example for LCD Drive Voltage Generation

## When an Internal Booster and Internal Operational Amplifiers are Used

To supply LCD drive voltage using the internal booster, circuits should be connected as shown in Figure 31. Here, contrast can be adjusted through the CT bits of the contrast-control instruction. Temperature can be compensated either through the CT bits or by controlling the reference voltage for the booster (Vci pin) using a thermistor.

Note that Vci is both a reference voltage and power supply for the booster; the reference voltage must therefore be adjusted using an emitter-follower or a similar element so that sufficient current can be supplied. In this case, Vci must be equal to or smaller than the $\mathrm{V}_{\mathrm{CC}}$ level.

The HD66717 incorporates a voltage-follower operational amplifier for each of V1 to V5 to reduce current flowing through the internal bleeder-resistors, which generate different levels of liquid-crystal drive voltages. Thus, potential differences between $\mathrm{V}_{\mathrm{CC}}$ and V 1 and between $\mathrm{V}_{\mathrm{EE}}$ and V 5 must be 0.4 V or greater. Note that the OPOFF pin must be grounded when using the operational amplifiers.


Figure 31 Internal Power Supply Circuit Example for LCD Drive Voltage Generation

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Figure 32 Temperature Compensation Circuit Example

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The HD66717's internal operational amplifiers have a reduced drive current to save current consumption; when the internal operational amplifiers cannot fully drive the LCD panel used, an appropriate capacitors must be inserted between each output of V1OUT to V5OUT and $\mathrm{V}_{\mathrm{CC}}$ to stabilize the operational amplifier output (Figure 33). Especially, the capacitors for V1OUT and V4OUT must be inserted when $1 / 26$ duty or $1 / 34$ duty drives.


Note: The capacitors for V1OUT and V4OUT must be inserted when $1 / 26$ duty or $1 / 34$ duty drives.
Figure 33 Operational Amplifier Output Stabilization Circuit Example

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## When an Internal Booster and External Bleeder-Resistors are Used

When the internal operational amplifiers cannot fully drive the LCD panel used, V1 to V5 voltages can be supplied through external bleeder-resistors (Figure 34). Here, the OPOFF pin must be set to the $\mathrm{V}_{\text {cc }}$ level to turn off the internal operational amplifiers. Since the internal contrast adjuster is disabled in this case, contrast must be adjusted externally. Double- and triple-boosters can be used as they are.


Note: 1. Resistance of each external bleeder resistor should be $5 \mathrm{k} \Omega$ to $15 \mathrm{k} \Omega$.
2. The bias current value for driving liquid-crystals can be varied by adjusting the resistance (2R) between the V2OUT and V3OUT pins.
3. The internal contrast-adjuster is disabled; contrast must be adjusted either by controlling the external variable resistor between $\mathrm{V}_{\mathrm{EE}}$ and V5OUT or Vci for the booster.
4. Vci is both a reference voltage and power supply for the booster; connect it to $\mathrm{V}_{\mathrm{CC}}$ directly or combine it with a transistor so that sufficient current can be obtained.
5. Vci must be smaller than $\mathrm{V}_{\mathrm{CC}}$.

Figure 34 External Bleeder-Resistor Example for LCD Drive Voltage Generation Power Supply Circuit

## Contrast Adjuster

## Multiplexing Drive System

Contrast for an LCD controlled by the multiplexing drive method can be adjusted by varying the liquidcrystal drive voltage (potential difference between $\mathrm{V}_{\mathrm{CC}}$ and V 5 ) through the CT bits of the contrast control instruction (electron volume function). See Figure 35 and Table 20. The value of a variable resistor (VR) can be adjusted within the range from 0.4 R through 6.4 R , where R is a reference resistance obtained by dividing the total resistance between $\mathrm{V}_{\mathrm{CC}}$ and V5.

The HD66717 incorporates a voltage-follower operational amplifier for each of V1 to V5 to reduce current flowing through the internal bleeder-resistors, which generate different levels of liquid-crystal drive voltages. Thus, potential differences between $\mathrm{V}_{\mathrm{CC}}$ and V 1 and between $\mathrm{V}_{\mathrm{EE}}$ and V 5 must be 0.4 V or greater. Note that the OPOFF pin must be grounded when using the operational amplifiers.

- $1 / 6$ bias (V2 and V3 pins left open)
- LCD drive voltage VLCD: $6 \mathrm{R} \times\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}\right) /(6 \mathrm{R}+\mathrm{VR})(\mathrm{VR}=$ a value within the range from 0.4 R to 6.4 R )
— VLCD adjustable range: $0.484 \times\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}\right) \leq \mathrm{VLCD} \leq 0.938 \times\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}\right)$
- Potential difference between $\mathrm{V}_{\mathrm{CC}}$ and $\mathrm{V} 1: \mathrm{R} \times\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}\right) /(6 \mathrm{R}+\mathrm{VR}) \geq 0.4(\mathrm{~V})$
— Potential difference between V5 and $\mathrm{V}_{\mathrm{EE}}: \mathrm{VR} \times\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}\right) /(6 \mathrm{R}+\mathrm{VR}) \geq 0.4(\mathrm{~V})$
- 1/4 bias (V2 and V3 pins short-circuited)
- LCD drive voltage VLCD: $4 \mathrm{R} \times\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}\right) /(4 \mathrm{R}+\mathrm{VR})(\mathrm{VR}=$ a value within the range from 0.4 R to 6.4 R )
- VLCD adjustable range: $0.385 \times\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}\right) \leq \mathrm{VLCD} \leq 0.909 \times\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}\right)$
- Potential difference between $\mathrm{V}_{\mathrm{CC}}$ and $\mathrm{V} 1: \mathrm{R} \times\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}\right) /(4 \mathrm{R}+\mathrm{VR})^{3} 0.4(\mathrm{~V})$
- Potential difference between V5 and $\mathrm{V}_{\mathrm{EE}}: \mathrm{VR} \times\left(\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}\right) /(4 \mathrm{R}+\mathrm{VR})^{3} 0.4(\mathrm{~V})$


## Static Drive System

Contrast for a statically-driven LCD, that is, annunciator display, can be adjusted through the AGND pin. The annunciators are driven statically by the potential difference between $\mathrm{V}_{\mathrm{CC}}$ and AGND. The AGND pin level must be equal to or greater than the GND level.

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Figure 35 Contrast Adjuster

Table 20 Contrast-Adjust Bits (CT) and Variable Resistor Values

|  | CT Register |  |  |  |
| :--- | :--- | :--- | :--- | :--- |
| CT3 | CT2 | CT1 | CTO | Variable Resistor Value (VR) |
| 0 | 0 | 0 | 0 | 6.4 R |
| 0 | 0 | 0 | 1 | 6.0 R |
| 0 | 0 | 1 | 0 | 5.6 R |
| 0 | 0 | 1 | 1 | 5.2 R |
| 0 | 1 | 0 | 0 | 4.8 R |
| 0 | 1 | 0 | 1 | 4.4 R |
| 0 | 1 | 1 | 0 | 4.0 R |
| 0 | 1 | 1 | 1 | 3.6 R |
| 1 | 0 | 0 | 0 | 3.2 R |
| 1 | 0 | 0 | 1 | 2.8 R |
| 1 | 0 | 1 | 0 | 2.4 R |
| 1 | 0 | 1 | 1 | 2.0 R |
| 1 | 1 | 0 | 0 | 1.6 R |
| 1 | 1 | 0 | 1 | 1.2 R |
| 1 | 1 | 1 | 0 | 0.8 R |
| 1 | 1 | 1 | 1 | 0.4 R |

## LCD Module Interface

Segment data output pins SEG1 to SEG60 can be connected either from left to right or right to left of an LCD panel according to the SFT pin level. When the SFT pin is grounded, SEG1 is connected to the far left of the panel, and when it is at the $\mathrm{V}_{\mathrm{CC}}$ level, SEG60 is connected to the far left. Either connection mode can be selected according to the LCD module layout and routing on a printed-circuit board. Figures 36 shows two examples.


Figure 36 LCD Module Interface Examples

## Segment Display and Annunciator Display

The HD66717 provides both segment display, which is driven by the multiplexing method, and annunciator display, which is driven statically. Annunciator display is driven at a logic operating voltage ( $\left.\mathrm{V}_{\mathrm{cC}}-\mathrm{AGND}\right)$ and is thus also available while the LCD drive power supply is turned off. Accordingly, annunciator display is suitable for displaying marks during system standby, when it is desirable to reduce current consumption. It is available in sleep mode, where internal multiplexing operations for character or segment display are halted. If an alternating signal is supplied to the EXM pin, it is also available in standby mode, where the internal R-C oscillator is halted. Here, AGND must be equal to or above the GND level.

Note that annunciator display cannot share character display drivers SEG and COM but require special drivers ASEG and ACOM that require long routing.

Tables 21 to 23 illustrates segment display and annunciator display.
Table 21 Comparison between Segment Display and Annunciator Display

| Item | Segment Display | Annunciator Display |
| :--- | :--- | :--- |
| Number of driven elements | 20 each by COMS1 and COMS2 | 10 |
| Blinking | Impossible | Possible |
| Segment drivers | SEG1-SEG60 <br> (shared with character display) | ASEG1-ASEG10 <br> (independent of character display) |
| Common drivers | COMS1, COMS2 | ACOM |
| LCD power supply | $\mathrm{V}_{\mathrm{cc}}-\mathrm{V} 5$ | $\mathrm{~V}_{\mathrm{cc}}-$ AGND |
|  | (LCD power supply necessary) | (LCD power supply unnecessary) |
| Normal mode display | Display possible together with <br> character display by multiplexing drive | Display possible by static drive |
|  | Impossible Possible by static drive <br> Sleep mode display (SEG and COM output $\mathrm{V}_{\mathrm{cc}}$ ) |  |
| Standby mode display | Impossible | Possible by supplying alternating |
| (without oscillation) | (SEG and COM output $\mathrm{V}_{\mathrm{cc}}$ ) | signal to the EXM pin |

Table 22 Correspondence between Segment Display SEGRAM Addresses (ASEG) and Driver Signals

| ASEG Address |  |  |  | Common Signal | Segment Signals |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MSB |  |  | LSB |  | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| 1 | 0 | 0 | 0 | COMS1 | SEG1/21/41 | SEG2/22/42 | SEG3/23/43 | SEG4/24/44 | SEG5/25/45 |
| 1 | 0 | 0 | 1 | COMS1 | SEG6/26/46 | SEG7/27/47 | SEG8/28/48 | SEG9/29/49 | SEG10/30/50 |
| 1 | 0 | 1 | 0 | COMS1 | SEG11/31/51 | SEG12/32/52 | SEG13/33/53 | SEG14/34/54 | SEG15/35/55 |
| 1 | 0 | 1 | 1 | COMS1 | SEG16/36/56 | SEG17/37/57 | SEG18/38/58 | SEG19/39/59 | SEG20/40/60 |
| 1 | 1 | 0 | 0 | COMS2 | SEG1/21/41 | SEG2/22/42 | SEG3/23/43 | SEG4/24/44 | SEG5/25/45 |
| 1 | 1 | 0 | 1 | COMS2 | SEG6/26/46 | SEG7/27/47 | SEG8/28/48 | SEG9/29/49 | SEG10/30/50 |
| 1 | 1 | 1 | 0 | COMS2 | SEG11/31/51 | SEG12/32/52 | SEG13/33/53 | SEG14/34/54 | SEG15/35/55 |
| 1 | 1 | 1 | 1 | COMS2 | SEG16/36/56 | SEG17/37/57 | SEG18/38/58 | SEG19/39/59 | SEG20/40/60 |

Table 23 Correspondence between Annunciator Display Addresses (AAN) and Driver Signals

|  | AAN Address |  |  |  | Common |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  | LSB | Signal |  | Bits 7, $\mathbf{6}$ | Bits 5, 4 | Bits 3, 2 | Bits 1, 0

Note: The annunciator is turned on when the corresponding even bit (bit $6,4,2$, or 0 ) is 1 , and the turned-on annunciator blinks when the corresponding odd bit (bit $7,5,3$, or 1 ) is 1 .

## Annunciator Drive

Figure 37 shows annunciator drive output waveforms in two modes.


Figure 37 Annunciator Drive Output Waveforms

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## Vertical Smooth Scroll

The HD66717 can scroll in the vertical direction in units of raster-rows. This function is achieved by writing character codes into the DDRAM area that is not being used for display. In other words, since the DDRAM corresponds to a 5-line $\times 12$-character display, one of the lines can be used to achieve continuous smooth vertical scroll even in a 4-line display. Here, after the fifth line is displayed, the first line is displayed again. Specifically, this function is controlled by incrementing or decrementing the value in the scroll-start line bits (SL2 to SL0) and scroll-start raster-row bits (SN2 to SN0) by 1. For example, to smoothly scroll up, first set SN2 to SN0 to 000, and increment SL2 to SL0 by 1 from 000 to 111 to scroll seven raster-rows. Then increment SN2 to SN0 to 001, and again increment SL2 to SL0 by 1 from 000 to 111 . To start displaying and scrolling from the first raster-row of the second line, update the first line of DDRAM data as desired during its non-display period.

Figure 38 shows an example of vertical smooth scrolling and Figure 39 shows an example of setting instructions for vertically scrolling upward in a 4-line display ( NL 1 and $\mathrm{NLO}=11$ ).


Figure 38 Example of Vertical Smooth Scrolling

| Set initial data to all DDRAM addresses |  | 9） 8 raster－row scrolled up$\begin{aligned} & \cdot \mathrm{SN} 2-0=001 \\ & \cdot \mathrm{SL} 2-0=000 \end{aligned}$ |  |
| :---: | :---: | :---: | :---: |
| 1）Not scrolled $\begin{aligned} & \cdot \text { SN2-0 }=000 \\ & \cdot \text { SL2-0 }=000 \end{aligned}$ | HITACHI LTD． <br> （0423－25－1111 <br> LCD Controll <br> Er ${ }^{-}$Driver |  |  |
| $\begin{aligned} & \text { 2) } 1 \text { raster-row scrolled } \\ & \text { up } \\ & \text {. SL2-0 }=001 \end{aligned}$ | HIIHLHI LID． 0423－25－1111 LCD Eontroll er Priver： | $\begin{aligned} & \text { 10) } 9 \text { raster-row scrolled } \\ & \text { up } \\ & . \text { SL2-0 }=001 \end{aligned}$ | 24425－25－1111 <br> LCD Controll <br> Er Dreiver <br> $>$ HDG6717 |
| 3） 2 raster－row scrolled up $\cdot S L 2-0=010$ | HIIHLHI LIL． 0423－25－1111 LCD Eontroll er br iver | 11） 10 raster－row scrolled up $\cdot \text { SL2-0 }=010$ | $124 \angle 3-\angle 0-11+1$ <br> LCD Eontroll <br> er Driver <br> $\geqslant$ HDG67！ |
| 4） 3 raster－row scrolled up $\cdot \text { SL2-0 }=011$ |  0423－25－1111 LCD Controll er priver い 1rw | 12） 11 raster－row scrolled up $\cdot S L 2-0=011$ |  <br> LCD Controll <br> e․ Dr iver <br> ＞）HD6E717 <br> เו－．ran．．．：－ |
| 5） 4 raster－row scrolled up $\text { - SL2-0 = } 100$ | $6423-25-111$ <br> LCD Eontroll <br>  <br> © unce717 | $\begin{aligned} & \text { 13) } 12 \text { raster-row scrolled } \\ & \text { up } \\ & \text {. SL2-0 }=100 \end{aligned}$ | LCD Controil <br> er Driver <br> ＞ H ［66717 <br> klen ！rimis．．． |
| $\begin{aligned} & \text { 6) } 5 \text { raster-row scrolled } \\ & \text { up } \\ & \text { SL2-0 }=101 \end{aligned}$ | $0423-25-111$ <br> LCD Eontroll <br> er Driver <br> S． Hike． 717 | 14） 13 raster－row scrolled up $\cdot S L 2-0=101$ | L゙CD Controil er Driver 7）HD66717 Na．．「ianifa |
| $\begin{aligned} & \text { 7) } 6 \text { raster-row scrolled } \\ & \text { up } \\ & \cdot \text { SL2-0 }=110 \end{aligned}$ | － $2425-25-111$ <br> LCD Eontroll <br> Er Driver <br> ＞）H | 15） 14 raster－row scrolled up $\text { - SL2-0 = } 110$ | God controil <br> er Driver <br> 3）HDG6717 <br> Helul Revirie |
| $\begin{aligned} & \text { 8) } 7 \text { raster-row scrolled } \\ & \text { up } \\ & \text { - SL2-0 }=111 \end{aligned}$ | 6423－25－1111 <br> LCD Controll <br> er Driver <br> ＞） HD 66717 | 16） 15 raster－row scrolled up $\cdot S L 2-0=111$ | LCD Controll er Dr iver ＞）HDGE717 HEw Device |

Figure 39 Example of Setting Instructions for Vertical Smooth Scroll
（4－line display（ NL 1 and $\mathrm{NLO}=11$ ）

## Line-Cursor Display

The HD66717 can assign a cursor attribute to an entire line corresponding to the address counter value by setting the LC bit to 1 (Table 24). One of three line-cursor modes can be selected: a black-white inverting blink cursor $(B / W=1)$, an underline cursor $(C=1)$, and a blink cursor $(B=1)$. The blink cycle for a black-white inverting blink cursor and for a blink cursor is 32 frames. These line-cursors are suitable for highlighting an index and/or marker, and for indicating an item in a menu with a cursor or an underline.

Figures 40 to 42 show three line-cursor examples.
Table 24 Address Counter Value and Line-Cursor

| Address Counter Value (AC) | Selected Line for Line-Cursor |
| :--- | :--- |
| 00 H to 0 BH | Entire 1st line (12 digits) |
| 10 H to 1 BH | Entire 2nd line (12 digits) |
| 20 H to 2 BH | Entire 3rd line (12 digits) |
| 30 H to 3 BH | Entire 4th line (12 digits) |
| 40 H to 4 BH | Entire 5th line (12 digits) |



Figure 40 Example of Black-White Inverting Blink Cursor (LC = 1; B/W = 1)


Figure 41 Example of Underline Cursor ( $\mathrm{LC}=1 ; \mathrm{C}=1$ )


Figure 42 Example of Blink Cursor ( $\mathbf{L C}=1 ; B=1$ )

## Double-Height Display

The HD66717 can double the height of any desired line from the first to third lines. A line can be selected by the DL3 to DL1 bits as listed in Table 25. All the standard font characters stored in the CGROM and CGRAM can be doubled in height, providing an easy-to-see display. Note that there should be no space between lines for double-height display (Figure 43).

Table 25 Double-Height Display Specifications

| DL3 | DL2 | DL1 | 2-Line Display <br> (NL1, NL0 = 01) | 3-Line Display <br> (NL1, NL0 = 10) | 4-Line Display <br> (NL1, NL0 = 11) |
| :--- | :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 1st \& 2nd lines: normal | 1st to 3rd lines: normal | 1st to 4th lines: normal |
| 0 | 0 | 1 | 1st line: double-height | 1st line: double-height <br> 2nd line: normal | 1st line: double-height <br> 2nd \& 3rd lines: normal |
| 0 | 1 | 0 | Disabled | 2nd line: double-height <br> 1st line: normal | 2nd line: double-height <br> 1st \& 3rd lines: normal |
| 0 | 1 | 1 | 1st line: double-height | Disabled | 1st \& 2nd lines: double-height |



Figure 43 Double-Height Display Examples

## Partial-Display-Off Function

The HD66717 can program the number of display lines (NL1 and NL0 bits), divide the internal operating frequency by four (OSC bit), and adjust the display contrast (CT bit). Combining these functions, the HD66717 can turn off the second and/or subsequent lines, displaying only the characters in the first line to reduce internal current consumption (partial-display-off function). This function is suitable for calendar or time display, which needs to be continuous during system standby with minimal current consumption. Here, the second to fourth non-displayed lines are constantly driven by the deselection level voltage, thus turning off the LCD for the lines.

Note that internal clock frequency is reduced to a quarter, quadrupling execution time of each instruction; MPU data transfer rate must be appropriately adjusted.

Table 26 lists partial-display-off function specifications and Figure 44 shows a sample display using the partial-display-off function.

## Table 26 Partial Display Off Function

| Function Item | Normal 4-Line Display | Partially-Off Display |
| :--- | :--- | :--- |
| Character display | 1st to 4th lines displayed | Only 1st line displayed |
| Segment display | Possible | Possible |
| Annunciator display | Possible | Possible |
| R-C oscillation frequency | 160 kHz | 160 kHz |
| Internal operating frequency | $160 \mathrm{kHz}(\mathrm{OSC} \mathrm{=0)}$ | $40 \mathrm{kHz}(\mathrm{OSC} \mathrm{=1)}$ |
| LCD single-line drive frequency | $2.7 \mathrm{kHz}(1 / 34$ duty ratio $)$ | $0.7 \mathrm{kHz} \mathrm{(1/10} \mathrm{duty} \mathrm{ratio)}$ |
| Frame frequency | 78 Hz | 66 Hz |

Note: Select an optimum LCD drive voltage (between $\mathrm{V}_{\mathrm{cc}}$ and V 5 ) for the multiplexing duty ratio used, using a reference voltage input pin (Vci) for the booster or the contrast-adjust bits (CT) .

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Figure 44 Example of Partially-Off Display (date and time indicated)

## Sleep Mode

Setting the sleep mode bit (SLP) to 1 puts the HD66717 in sleep mode, where the device halts all the internal display operations except for annunciator display operations, thus reducing current consumption. Specifically, character and segment displays, which are controlled by the multiplexing drive method, are completely halted. Here, all the SEG (SEG1 to SEG60) and COM (COM1 to COM34) pins output the $\mathrm{V}_{\mathrm{CC}}$ level, resulting in no display. If the AMP bit is set to 0 in sleep mode, the LCD drive power supply can be turned off, reducing the total current consumption of the LCD module.

Annunciators can be normally displayed in sleep mode. Since they are driven at logic operating power supply voltage ( $\mathrm{V}_{\mathrm{CC}}-\mathrm{AGND}$ ), they are available even if the LCD power supply is turned off (AMP $=0$ ). This function allows time and alarm marker indication during system standby with reduced current consumption.

During sleep mode, no instructions can be accepted for character/segment display and neither DDRAM, CGRAM, nor SEGRAM can be accessed.

Table 27 compares the functions of sleep mode and standby mode.
Table 27 Comparison of Sleep Mode and Standby Mode

| Function Item | Sleep Mode (SLP =1) | Standby Mode (STB = 1) |
| :--- | :--- | :--- |
| Character display | Turned off | Turned off |
| Segment display | Turned off | Turned off |
| Annunciator display | Can be turned on | Can be turned on when an alternating <br> signal is supplied to the EXM pin |
| R-C oscillation | Normally operates | Halted |

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## Standby Mode

Setting the standby mode bit (STB) to 1 puts the HD66717 in standby mode, where the device stops completely, halting all internal operations including the R-C oscillator, thus further reducing current consumption compared to that in sleep mode. Specifically, character and segment displays, which are controlled by the multiplexing drive method, are completely halted. Here, all the SEG (SEG1 to SEG60) and COM (COM4 to COM34) pins output the $\mathrm{V}_{\mathrm{CC}}$ level, resulting in no display. If the AMP bit is set to 0 in standby mode, the LCD drive power supply can be turned off.

Annunciators can be displayed simply by supplying an approximately $40-\mathrm{Hz}$ alternating signal for the LCD drive signals to the EXM pin externally. If annunciator display is unnecessary during standby mode, the EXM pin must be fixed to the $\mathrm{V}_{\mathrm{CC}}$ or GND level and the annunciator display-on bit (DA) set to 0 .

During standby mode, no instructions can be accepted other than those for annunciator display and the start-oscillator instruction. To cancel standby mode, issue the start-oscillator instruction to stabilize R-C oscillation before setting the STB bit to 0 .

Figure 45 shows the procedure for setting and cancelling standby mode.


Figure 45 Procedure for Setting and Cancelling Standby Mode

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## Absolute Maximum Ratings*

| Item | Symbol | Unit | Value | Notes |
| :--- | :--- | :--- | :--- | :--- |
| Power supply voltage (1) | $\mathrm{V}_{\mathrm{CC}}$ | V | -0.3 to +7.0 | 1 |
| Power supply voltage (2) | $\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{EE}}$ | V | -0.3 to +15.0 | 1,2 |
| Input voltage | Vt | V | -0.3 to $\mathrm{V}_{\mathrm{CC}}+0.3$ | 1 |
| Operating temperature | $\mathrm{T}_{\text {opr }}$ | ${ }^{\circ} \mathrm{C}$ | -20 to +75 |  |
| Storage temperature | $\mathrm{T}_{\text {stg }}$ | ${ }^{\circ} \mathrm{C}$ | -40 to +125 | 4 |

Note: * If the LSI is used above these absolute maximum ratings, it may become permanently damaged. Using the LSI within the following electrical characteristic limits is strongly recommended for normal operation. If these electrical characteristic conditions are also exceeded, the LSI will malfunction and cause poor reliability.

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DC Characteristics $\left(\mathrm{V}_{\mathrm{cC}}=2.4 \mathrm{~V}\right.$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=-\mathbf{2 0}$ to $\left.+75^{\circ} \mathrm{C}^{*}{ }^{\mathbf{3}}\right)$

| Item | Symbol | Min | Typ | Max | Unit | Test Condition | Notes* |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input high voltage | VIH | $0.7 \mathrm{~V}_{\mathrm{cc}}$ | - | $\mathrm{V}_{\text {cc }}$ | V |  | 6 |
| Input low voltage | VIL | -0.3 | - | $0.15 \mathrm{~V}_{\mathrm{cc}}$ | V | $\mathrm{V}_{\mathrm{cc}}=2.4$ to 3.0 V | 6 |
|  |  | -0.3 | - | 0.6 | V | $\mathrm{V}_{\mathrm{cc}}=3.0$ to 5.5 V |  |
| Output high voltage <br> (1) (DB0-DB7 pins) | VOH1 | $0.75 \mathrm{~V}_{\text {cc }}$ | - | - | V | $\mathrm{I}_{\text {OH }}=-0.1 \mathrm{~mA}$ |  |
| Output low voltage (1) (DB0-DB7 pins) | VOL1 | - | - | $0.2 \mathrm{~V}_{\mathrm{cc}}$ | V | $\mathrm{I}_{\mathrm{oL}}=0.1 \mathrm{~mA}$ |  |
| Output low voltage (2) (SDA pin) | VOL2 | - | - | $0.2 \mathrm{~V}_{\mathrm{cc}}$ | V | $\begin{aligned} & \mathrm{V}_{\mathrm{cc}}=2.4 \text { to } 4.5 \mathrm{~V} \\ & \mathrm{I}_{\mathrm{oL}}=0.4 \mathrm{~mA} \end{aligned}$ | 7 |
|  |  | - | - | 0.4 | V | $\begin{aligned} & \hline \mathrm{V}_{\mathrm{cc}}=4.5 \text { to } 5.5 \mathrm{~V} \\ & \mathrm{I}_{\mathrm{oc}}=1.0 \mathrm{~mA} \end{aligned}$ |  |
| Driver ON resistance (COM pins) | $\mathrm{R}_{\text {com }}$ | - | 2 | 20 | k $\Omega$ | $\begin{aligned} & \pm \mathrm{Id}=0.05 \mathrm{~mA}(\mathrm{COM}) \\ & \mathrm{VLCD}=4 \mathrm{~V} \end{aligned}$ | 8 |
| Driver ON resistance (SEG pins) | $\mathrm{R}_{\text {sEG }}$ | - | 2 | 30 | k $\Omega$ | $\begin{aligned} & \pm \mathrm{Id}=0.05 \mathrm{~mA}(\mathrm{SEG}) \\ & \mathrm{VLCD}=4 \mathrm{~V} \end{aligned}$ | 8 |
| I/O leakage current | $\mathrm{I}_{\mathrm{u}}$ | -1 | - | 1 | $\mu \mathrm{A}$ | $\mathrm{V}_{\mathrm{N}}=0$ to $\mathrm{V}_{\mathrm{cc}}$ | 9 |
| Pull-up MOS current (RESET* ${ }^{\text {pin }}$ | -Ip | 10 | 50 | 120 | $\mu \mathrm{A}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{cc}}=3 \mathrm{~V} \\ & \mathrm{Vin}=0 \mathrm{~V} \end{aligned}$ |  |
| Current consumption during normal operation ( $\mathrm{V}_{\mathrm{cc}}$-GND) | $\mathrm{I}_{\text {op }}$ | - | 30 | 60 | $\mu \mathrm{A}$ | $\mathrm{R}_{\mathrm{f}}$ oscillation, external clock, $\mathrm{V}_{\mathrm{cc}}=3 \mathrm{~V}, \mathrm{f}_{\text {osc }}=160 \mathrm{kHz}$, 1/34 duty | 10, 11 |
| Current consumption during sleep mode ( $\mathrm{V}_{\mathrm{cc}}$-GND) | $\mathrm{I}_{\text {sL }}$ | - | 25 | - | $\mu \mathrm{A}$ | $\mathrm{R}_{\mathrm{f}}$ oscillation, external clock, $\mathrm{V}_{\mathrm{cc}}=3 \mathrm{~V}, \mathrm{f}_{\mathrm{osc}}=160 \mathrm{kHz}$ | 10,11 |
| Current consumption during standby mode ( $\mathrm{V}_{\mathrm{cc}}$-GND) | $\mathrm{I}_{\text {st }}$ | - | 0.1 | 5 | $\mu \mathrm{A}$ | No $\mathrm{R}_{\mathrm{f}}$ oscillation, $\mathrm{V}_{\mathrm{cc}}=3 \mathrm{~V}, \mathrm{Ta}=25^{\circ} \mathrm{C}$ | 10,11 |
| LCD power supply current $\left(\mathrm{V}_{\mathrm{Cc}}-\mathrm{V}_{\mathrm{EE}}\right)$ | $\mathrm{I}_{\text {EE }}$ | - | 25 | 60 | $\mu \mathrm{A}$ | $\begin{aligned} & \mathrm{V}_{\text {cc }}-\mathrm{V}_{\text {EE }}=8 \mathrm{~V}, \\ & \mathrm{f}_{\text {osc }}=160 \mathrm{kHz} \end{aligned}$ <br> VREF-VREFM: shortcircuited |  |
| LCD voltage with $1 / 4$ bias $\left(V_{C C}-V_{E E}\right)$ | VLCD1 | 3.0 | - | 13.0 | v | V2-V3 short-circuited | 12 |
| LCD voltage with $1 / 6$ bias $\left(V_{\mathrm{CC}}-V_{\mathrm{EE}}\right)$ | VLCD2 | 3.0 | - | 13.0 | v | V2-V3 open | 12 |

Note: * Refer to the Electrical Characteristics Notes section following these tables.

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## Booster Characteristics

| Item | Symbol | Min | Typ | Max | Unit | Test Condition | Notes* |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Output voltage (V5OUT2 pin) | VUP2 | 8.0 | 8.8 | - | V | $\begin{aligned} & \mathrm{V}_{\mathrm{cc}}=\mathrm{Vci}=4.5 \mathrm{~V}, \\ & \mathrm{I}_{0}=0.1 \mathrm{~mA}, \mathrm{C}=1 \mu \mathrm{~F}, \\ & \mathrm{f}_{\mathrm{osc}}=160 \mathrm{kHz}, \mathrm{~T}_{\mathrm{a}}=25^{\circ} \mathrm{C} \end{aligned}$ | 15 |
| Output voltage (V50UT3 pin) | VUP3 | 7.0 | 7.9 | - | V | $\begin{aligned} & \mathrm{V}_{\mathrm{cc}}=\mathrm{Vci}=2.7 \mathrm{~V}, \\ & \mathrm{I}_{0}=0.1 \mathrm{~mA}, \mathrm{C}=1 \mu \mathrm{~F}, \\ & \mathrm{f}_{\mathrm{osc}}=160 \mathrm{kHz}, \mathrm{~T}_{\mathrm{a}}=25^{\circ} \mathrm{C} \end{aligned}$ | 15 |
| Input voltage | VCi | 1.0 | - | 5.0 | V | $\mathrm{Vci}^{2} \mathrm{~V}_{\mathrm{cc}}$ | 15 |

Note: * Refer to the Electrical Characteristics Notes section following these tables.

AC Characteristics $\left(\mathrm{V}_{\mathrm{CC}}=\mathbf{2 . 4 V}\right.$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=\mathbf{- 2 0}$ to $\left.+75^{\circ} \mathrm{C}^{*}{ }^{\mathbf{3}}\right)$

Clock Characteristics ( $\mathrm{V}_{\mathrm{CC}}=2.4 \mathrm{~V}$ to 5.5 V )

| Item |  | Symbo | Min | Typ | Max | Unit | Test Condition | Notes* |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| External clock operation | External clock frequency | $\mathrm{f}_{\text {cp }}$ | 20 | 160 | 350 | kHz |  | 13 |
|  | External clock duty ratio | Duty | 45 | 50 | 55 | \% |  |  |
|  | External clock rise time | $\mathrm{t}_{\text {rep }}$ | - | - | 0.2 | $\mu \mathrm{s}$ |  |  |
|  | External clock fall time | $\mathrm{t}_{\text {cip }}$ | - | - | 0.2 | $\mu \mathrm{s}$ |  |  |
| $\mathrm{R}_{\mathrm{t}}$ oscillation | Clock oscillation frequency | $\mathrm{f}_{\text {osc }}$ | 120 | 160 | 200 | kHz | $\begin{aligned} & R_{\mathrm{f}}=150 \mathrm{k} \Omega, \\ & \mathrm{~V}_{\mathrm{cc}}=3 \mathrm{~V} \end{aligned}$ | 14 |

Note: * Refer to the Electrical Characteristics Notes section following these tables.

Read \& Write Bus Interface Timing Characteristics with Read Operation ( $\mathrm{V}_{\mathrm{CC}}=\mathbf{2 . 4 V}$ to 4.5V)

| Item | Symbo | Min | Typ | Max | Unit | Test Condition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Enable cycle time | $\mathrm{t}_{\text {coce }}$ | 1000 | - | - | ns | Figures 52 and 53 |
| Enable pulse width (high level) | $\mathrm{PW}_{\text {EH }}$ | 450 | - | - |  |  |
| Enable rise/fall time | $\mathrm{t}_{\mathrm{E},}, \mathrm{t}_{\mathrm{E}}$ | - | - | 25 |  |  |
| Address set-up time (RS, R/W to E) | $\mathrm{t}_{\text {As }}$ | 60 | - | - |  |  |
| Address hold time | $\mathrm{t}_{\text {AH }}$ | 20 | - | - |  |  |
| Data set-up time | $\mathrm{t}_{\text {osw }}$ | 195 | - | - |  |  |
| Data hold time | $\mathrm{t}_{\mathrm{H}}$ | 30 | - | - |  |  |
| Read data delay time | $\mathrm{t}_{\text {DOR }}$ | - | - | 400 | ns | Figure 53 |
| Read data hold time | $\mathrm{t}_{\text {онR }}$ | 5 | - | - |  |  |

Read \& Write Bus Interface Timing Characteristics with Read Operation ( $\mathrm{V}_{\mathrm{CC}}=4.5 \mathrm{~V}$ to 5.5 V )

| Item | Symbo | Min | Typ | Max | Unit | Test Condition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Enable cycle time | $\mathrm{t}_{\text {crce }}$ | 500 | - | - | ns | Figures 52 and 53 |
| Enable pulse width (high level) | $\mathrm{PW}_{\text {EH }}$ | 230 | - | - |  |  |
| Enable rise/fall time | $\mathrm{t}_{\mathrm{E},}, \mathrm{t}_{\mathrm{E},}$ | - | - | 20 |  |  |
| Address set-up time (RS, R/W to E) | $\mathrm{t}_{\text {As }}$ | 40 | - | - |  |  |
| Address hold time | $\mathrm{t}_{\text {AH }}$ | 30 | - | - |  |  |
| Data set-up time | $\mathrm{t}_{\text {osw }}$ | 80 | - | - |  |  |
| Data hold time | $\mathrm{t}_{\mathrm{H}}$ | 30 | - | - |  |  |
| Read data delay time | $\mathrm{t}_{\text {DOR }}$ | - | - | 200 | ns | Figure 53 |
| Read data hold time | $\mathrm{t}_{\text {онR }}$ | 5 | - | - |  |  |

Write Bus Interface Timing Characteristics without Read Operation ( $\mathrm{V}_{\mathrm{CC}}=\mathbf{2 . 4 V}$ to 5.5 V )

| Item | Symbol | Min | Typ | Max | Unit | Test Condition |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Enable cycle time |  | $\mathrm{t}_{\mathrm{CYCE}}$ | 500 | - | - | ns |
| Enable pulse width | $\mathrm{V}_{\mathrm{CC}}=2.4$ to 3.0 V | $\mathrm{P}_{\mathrm{WEH}}$ | 200 | - | - | Figure 52 |
| ("High" level) | $\mathrm{V}_{\mathrm{CC}}=3.0$ to 5.5 V | $\mathrm{P}_{\mathrm{wEH}}$ | 150 | - | - |  |
| Enable rise/fall time | $\mathrm{t}_{\mathrm{Er}}, \mathrm{t}_{\mathrm{Ef}}$ | - | - | 20 |  |  |
| Address set-up time (RS, R/W to E$)$ | $\mathrm{t}_{\mathrm{AS}}$ | 60 | - | - |  |  |
| Address hold time | $\mathrm{t}_{\mathrm{AH}}$ | 20 | - | - |  |  |
| Data set-up time | $\mathrm{t}_{\mathrm{DSW}}$ | 140 | - | - |  |  |
| Data hold time | $\mathrm{t}_{\mathrm{H}}$ | 30 | - | - |  |  |

Clock-Synchronized Serial Interface Operation ( $\mathrm{V}_{\mathrm{CC}}=2.4 \mathrm{~V}$ to 5.5 V )

| Item | Symbol | Min | Typ | Max | Unit | Test Condition |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Serial clock cycle time | $\mathrm{t}_{\mathrm{sCYC}}$ | 1 | - | 20 | $\mu \mathrm{~s}$ | Figure 54 |
| Serial clock high-level width | $\mathrm{t}_{\mathrm{sCH}}$ | 400 | - | - | ns |  |
| Serial clock low-level width | $\mathrm{t}_{\mathrm{sCL}}$ | 400 | - | - |  |  |
| Serial clock rise/fall time | $\mathrm{t}_{\mathrm{sc},}, \mathrm{t}_{\mathrm{sCf}}$ | - | - | 50 |  |  |
| Chip select set-up time | $\mathrm{t}_{\mathrm{cSU}}$ | 60 | - | - |  |  |
| Chip select hold time | $\mathrm{t}_{\mathrm{CH}}$ | 200 | - | - |  |  |
| Serial input data set-up time | $\mathrm{t}_{\mathrm{sISU}}$ | 200 | - | - |  |  |
| Serial input data hold time | $\mathrm{t}_{\mathrm{sIH}}$ | 200 | - | - |  |  |

$I^{2} \mathrm{C}$ bus Interface Operation $\left(\mathrm{V}_{\mathrm{CC}}=2.4 \mathrm{~V}\right.$ to $\left.\mathbf{4 . 5 V}\right)$

| Item | Symbol | Min | Typ | Max | Unit | Test Condition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SCL clock cycle time | $\mathrm{t}_{\text {scL }}$ | 2 | - | 20 | $\mu \mathrm{s}$ | Figure 55 |
| SCL clock high-level width | $\mathrm{t}_{\text {SCLH }}$ | 500 | - | - | ns |  |
| SCL clock low-level width | $\mathrm{t}_{\text {sclu }}$ | 1000 | - | - |  |  |
| SCL/SDA rise/fall time | $\mathrm{t}_{\mathrm{s} \text {, }}, \mathrm{t}_{\text {st }}$ | - | - | 300 |  |  |
| Bus free time | $\mathrm{t}_{\text {BUF }}$ | 100 | - | - |  |  |
| Start hold time | $\mathrm{t}_{\text {STAH }}$ | 500 | - | - |  |  |
| Retransmit start set-up time | $\mathrm{t}_{\text {STAS }}$ | 500 | - | - |  |  |
| Stop set-up time | $\mathrm{t}_{\text {stos }}$ | 500 | - | - |  |  |
| SDA data set-up time | $\mathrm{t}_{\text {SDAS }}$ | 140 | - | - |  |  |
| SDA data hold time | $\mathrm{t}_{\text {SDAH }}$ | 0 | - | - |  |  |

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I2C bus Interface Operation ( $\mathrm{V}_{\mathrm{CC}}=4.5 \mathrm{~V}$ to 5.5 V )

| Item | Symbol | Min | Typ | Max | Unit | Test Condition |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| SCL clock cycle time | $\mathrm{t}_{\mathrm{SCL}}$ | 2 | - | 20 | $\mu \mathrm{~s}$ | Figure 55 |
| SCL clock high-level width | $\mathrm{t}_{\mathrm{SCLH}}$ | 500 | - | - | ns |  |
| SCL clock low-level width | $\mathrm{t}_{\mathrm{sCLL}}$ | 1000 | - | - |  |  |
| SCL/SDA rise/fall time | $\mathrm{t}_{\mathrm{st}}, \mathrm{t}_{\mathrm{St}}$ | - | - | 300 |  |  |
| Bus free time | $\mathrm{t}_{\mathrm{BUF}}$ | 100 | - | - |  |  |
| Start hold time | $\mathrm{t}_{\mathrm{STAH}}$ | 500 | - | - |  |  |
| Retransmit start set-up time | $\mathrm{t}_{\mathrm{STAS}}$ | 500 | - | - |  |  |
| Stop set-up time | $\mathrm{t}_{\text {STOS }}$ | 500 | - | - |  |  |
| SDA data set-up time | $\mathrm{t}_{\text {SDAS }}$ | 100 | - | - |  |  |
| SDA data hold time | $\mathrm{t}_{\text {SDAH }}$ | 0 | - | - |  |  |

Reset Timing ( $\mathrm{V}_{\mathrm{CC}}=2.4 \mathrm{~V}$ to 5.5 V )

| Item | Symbol | Min | Typ | Max | Unit | Test Condition |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Reset low-level width | $\mathrm{t}_{\text {RES }}$ | 10 | - | - | ms | Figure 56 |

## Electrical Characteristics Notes

1. All voltage values are referred to $\mathrm{GND}=0 \mathrm{~V}$. If the LSI is used above the absolute maximum ratings, it may become permanently damaged. Using the LSI within the given electrical characteristic is strongly recommended to ensure normal operation. If these electrical characteristic are exceeded, the LSI may malfunction or exhibit poor reliability.
2. $\mathrm{V}_{\mathrm{CC}} \geq \mathrm{V} 1 \geq \mathrm{V} 2 \geq \mathrm{V} 3 \geq \mathrm{V} 4 \geq \mathrm{V} 5>\mathrm{V}_{\mathrm{EE}}$ must be maintained.
3. For die products, specified up to $75^{\circ} \mathrm{C}$.
4. For die products, specified by the common die shipment specification.
5. The following four circuits are I/O pin configurations except for liquid crystal display output (Figure 46).


Figure 46 I/O Pin Configurations

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6. The TEST pin must be grounded and the ID5 to ID0, IM1, IM0, SFT, EXM, and OPOFF pins must be grounded or connected to $\mathrm{V}_{\mathrm{CC}}$.
7. Applies to the ACK bit for $\mathrm{I}^{2} \mathrm{C}$ bus interface.
8. Applies to resistor values (RCOM) between power supply pins $\mathrm{V}_{\mathrm{CC}}$, V1OUT, V4OUT, V5OUT and common signal pins (COM1 to COM32, COMS1, and COMS2), and resistor values (RSEG) between power supply pins $\mathrm{V}_{\mathrm{CC}}$, V2OUT, V3OUT, V5OUT and segment signal pins (SEG1 to SEG60).
9. This excludes the current flowing through pull-up MOSs and output drive MOSs.
10. This excludes the current flowing through the input/output units. The input level must be fixed high or low because through current increases if the CMOS input is left floating.
11. The following shows the relationship between the operation frequency ( $\mathrm{f}_{\mathrm{OSC}}$ ) and current consumption ( $\mathrm{I}_{\mathrm{CC}}$ ) (Figure 47).


Figure 47 Relationship between the Operation Frequency and Current Consumption
12. Each COM and SEG output voltage is within $\pm 0.15 \mathrm{~V}$ of the LCD voltage $\left(\mathrm{V}_{\mathrm{CC}}, \mathrm{V} 1, \mathrm{~V} 2, \mathrm{~V} 3, \mathrm{~V} 4, \mathrm{~V} 5\right)$ when there is no load.
13. Applies to the external clock input (Figure 48).


Figure 48 External Clock Supply
14. Applies to the internal oscillator operations using oscillation resistor $R_{f}$ (Figure 49).


Figure 49 Internal Oscillation
15. Booster characteristics test circuits are shown in Figure 50.


Figure 50 Booster

Referential data
VUP2 $=\mathrm{V}_{\mathrm{CC}}-$ V5OUT2 $\quad$ VUP3 $=\mathrm{V}_{\mathrm{CC}}-$ V5OUT3
(i) Relationship between the obtained voltage and input voltage

(ii) Relationship between the obtained voltage and temperature


$$
\begin{aligned}
& \mathrm{Vci}=\mathrm{V}_{\mathrm{CC}}=4.5 \mathrm{~V}, \mathrm{R}_{\mathrm{f}}=180 \mathrm{k} \Omega, \\
& \mathrm{I}_{\mathrm{O}}=0.1 \mathrm{~mA}
\end{aligned}
$$


$\mathrm{Vci}=\mathrm{V}_{\mathrm{CC}}=2.7 \mathrm{~V}, \mathrm{R}_{\mathrm{f}}=150 \mathrm{k} \Omega$, $\mathrm{I}_{\mathrm{O}}=0.1 \mathrm{~mA}$
(iii) Relationship between the obtained voltage and capacitance


Figure 50 Booster (cont)

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(iv) Relationship between the obtained voltage and the load current.

$\mathrm{Vci}=\mathrm{V}_{\mathrm{CC}}=4.5 \mathrm{~V}, \mathrm{Rf}=180 \mathrm{k} \Omega$
$\mathrm{Ta}=25^{\circ} \mathrm{C}$


$$
\begin{gathered}
\mathrm{Vci}=\mathrm{V}_{\mathrm{CC}}=2.7 \mathrm{~V}, \mathrm{Rf}=150 \mathrm{k} \Omega \\
\mathrm{Ta}=25^{\circ} \mathrm{C}
\end{gathered}
$$

Figure 50 Booster (cont)

## Load Circuits

## AC Characteristics Test Load Circuits

| Data bus : SDA |
| :---: | :---: |
| Test Pointo |

Figure 51 Load Circuit

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## Timing Characteristics



Figure 52 Bus Write Operation


Figure 53 Bus Read Operation

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Figure 54 Clock-Synchronized Serial Interface Timing


Figure $55 \quad I^{2} \mathbf{C}$ Bus Interface Timing


Figure 56 Reset Timing

